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6.641 Electromagnetic Fields, Forces, and Motion  
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## Final Review Packet

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## Practice Problems

### Problem 1

A perfectly-conducting channel of height  $D$  and width  $2W$  is divided into two free-space regions by a very thin perfectly-conducting slide as shown below. The  $\hat{z}$ -directed magnetic fields in the left and right regions are  $H_l(t)\hat{z}$  and  $H_r(t)\hat{z}$ , respectively. The slide has a mass  $M$  per unit length in the  $\hat{z}$  direction, has a variable position  $\xi(t)$ , and makes a frictionless but perfectly-conducting contact with the channel.

(A) Assume that at  $t = 0$ ,  $H_l = H_{l_0}$ ,  $H_r = H_{r_0}$ , and  $\xi = \xi_0$ . Determine  $H_l(t)$  and  $H_r(t)$  in terms of  $H_{l_0}$ ,  $H_{r_0}$ ,  $\xi_0$ , and  $\xi(t)$ .

(B) The very thin slide supports a surface current  $K\hat{y}$  which separates  $H_l\hat{z}$  and  $H_r\hat{z}$ . This current interacts with the neighboring magnetic fields to produce a force  $F\hat{x}$  on the slide per unit length in the  $\hat{z}$  direction. Evaluate  $F$ .

(C) Use the results of (B) to find a function  $V(\xi)$  so that

$$\frac{d}{dt} \left( \frac{M}{2} \left( \frac{d\xi(t)}{dt} \right)^2 + V(\xi) \right) = 0.$$

(D) At  $t = 0$ ,  $\frac{d\xi}{dt} = 0$ . Determine the velocity of the slide when it first reaches  $\xi(t) = 0$ .

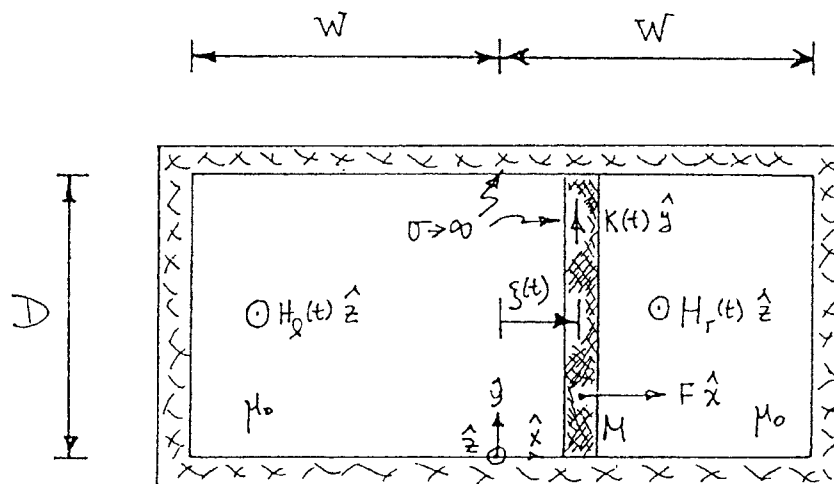


Figure 1: A perfectly-conducting channel with a perfectly conducting slide.

## Problem 2

A conducting plate of thickness  $\Delta$ , permeability  $\mu_0$  and conductivity  $\sigma$  moves with velocity  $U$  in the  $\hat{z}$  direction between two perfectly-permeable plates as shown below. At  $z = 0$ ,  $\bar{B}$  is constrained to be  $\text{Re}\{B_0 e^{j\omega t}\} \hat{x}$  by exciter coils. The perfectly-permeable plates confine  $\bar{B}$  to the region  $0 \leq z \leq L$  so that  $\bar{B} = 0 \hat{x}$  at  $z = L$ . Within the region  $0 \leq z \leq L$ ,  $\bar{B}$  is approximated by  $\bar{B} = B_x(z, t) \hat{x}$ . Ignore fringing fields.

- (A) Derive a differential equation for  $B_x$  in the conducting slab for  $0 \leq z \leq L$ .
- (B) Determine  $B_x$  for  $0 \leq z \leq L$ .
- (C) Determine the external force  $f$  needed to pull the plate in the  $\hat{z}$  direction at velocity  $U$ .

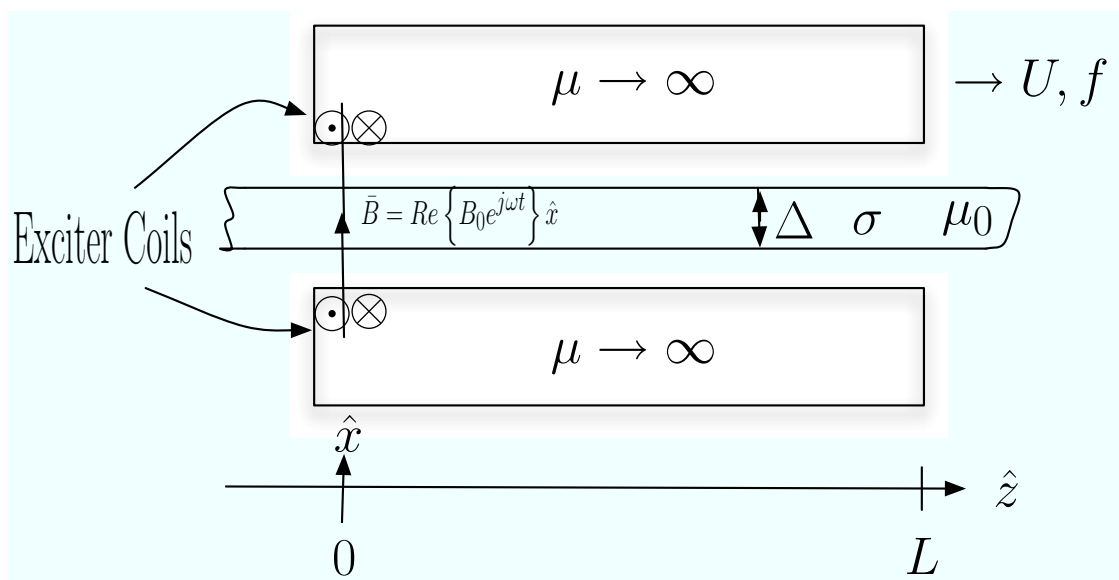


Figure 2: A conducting plate moving between infinitely permeable plates. (Image by MIT OpenCourseWare.)

**Problem 3**

A simplified model of a Van de Graaff generator is shown below. A belt with permittivity  $\epsilon$ , conductivity  $\sigma$ , thickness  $\delta$ , and width  $W$  travels to the right with velocity  $U$ . At  $x = 0$ , a charge source maintains the charge density  $\rho_0$  on the belt. At  $x = l$ , a charge collector collects all the charge off the belt. An external resistance  $R$  is connected from the charge collector to the charge source. Determine the current  $i$  through the resistor when the generator is in steady state.

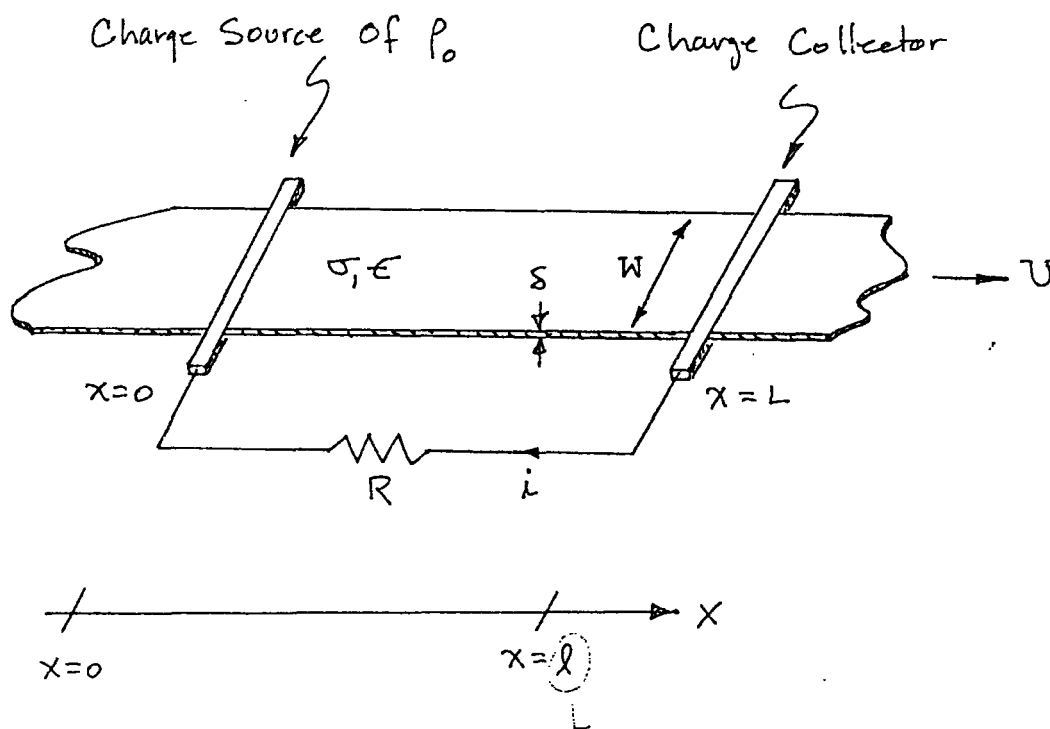


Figure 3: A simplified Van de Graaff generator.

**Problem 4**

A pair of grounded perfectly-conducting plates of infinite extent in the  $\hat{y}$  and  $\hat{z}$  directions are located at  $x = \Delta$  and  $x = -\Delta$  as shown below. A fluid having permittivity  $\epsilon$  and conductivity  $\sigma$  flows with uniform velocity  $U$  in the  $\hat{z}$  direction between the plates. At  $t = 0$ , the fluid has a charge distribution given by

$$\rho(x, y, z) = \begin{cases} \rho_0 \sin\left(\frac{\pi x}{\Delta}\right) e^{-ky^2} & |z| \leq \delta \\ 0 & |z| > \delta \end{cases}$$

Determine  $\rho(x, y, z, t)$  between the plates for  $t > 0$ ,  $|x| \leq \Delta$ , and all  $y$  and  $z$ .

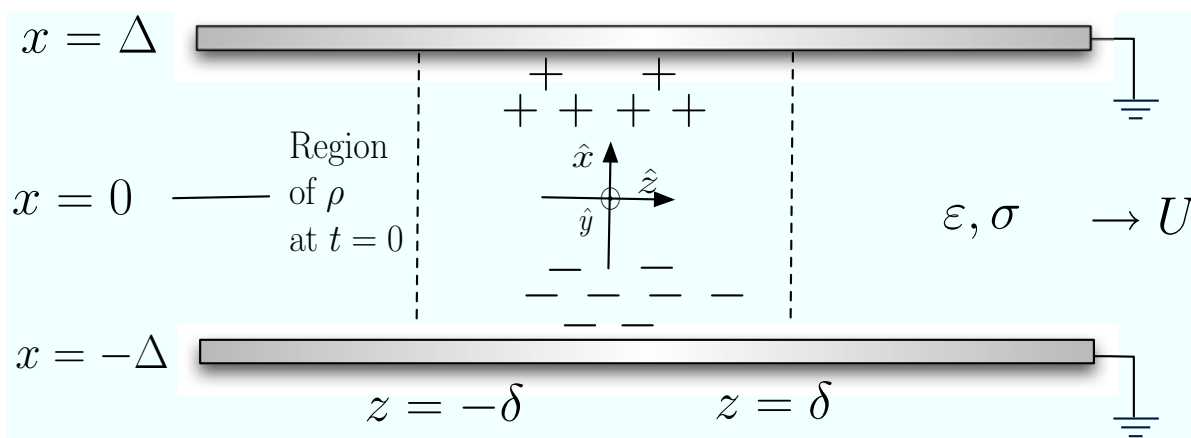


Figure 4: A pair of grounded perfectly-conducting plates enclose a moving conductor (Image by MIT OpenCourseWare.)

**Problem 5**

A thin sheet having effective surface conductivity  $\sigma_0$  moves in the  $\hat{z}$  direction with velocity  $U$  as shown below. The sheet is symmetrically located at a distance  $\Delta$  between two potential sources. The potential sources are symmetrically excited as traveling waves with frequency  $\omega$  and wave number  $k$ . Assume  $k\Delta \ll 1$  and make appropriate assumptions.

- (A) Find the electric field components  $E_x$  and  $E_z$  just above and below the thin sheet.
- (B) Find the free surface charge in the thin sheet.
- (C) Find the spatially and temporally averaged  $\hat{z}$ -directed force per  $y - z$  area which acts on the sheet.

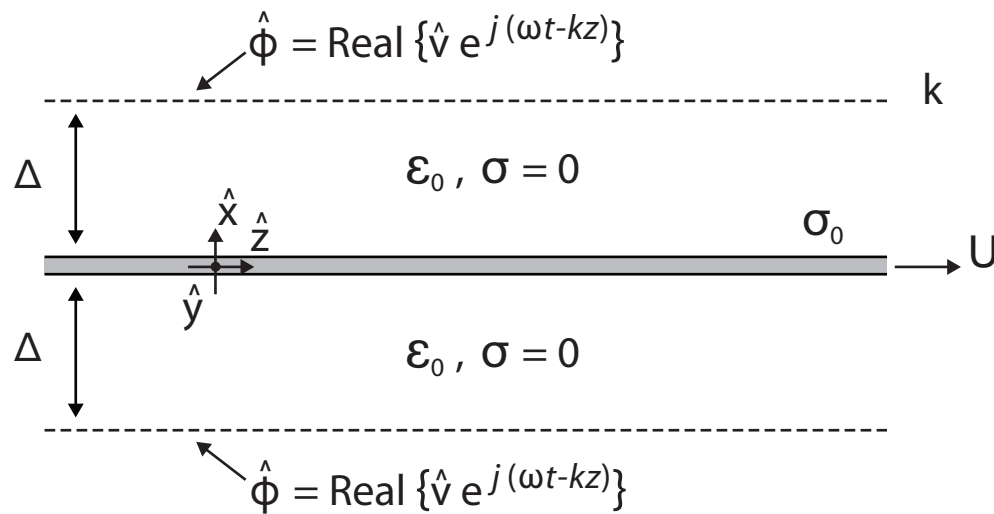


Figure 5: A thin sheet (Image by MIT OpenCourseWare.)

## Final Exam 1998 Solutions

## Problem 1

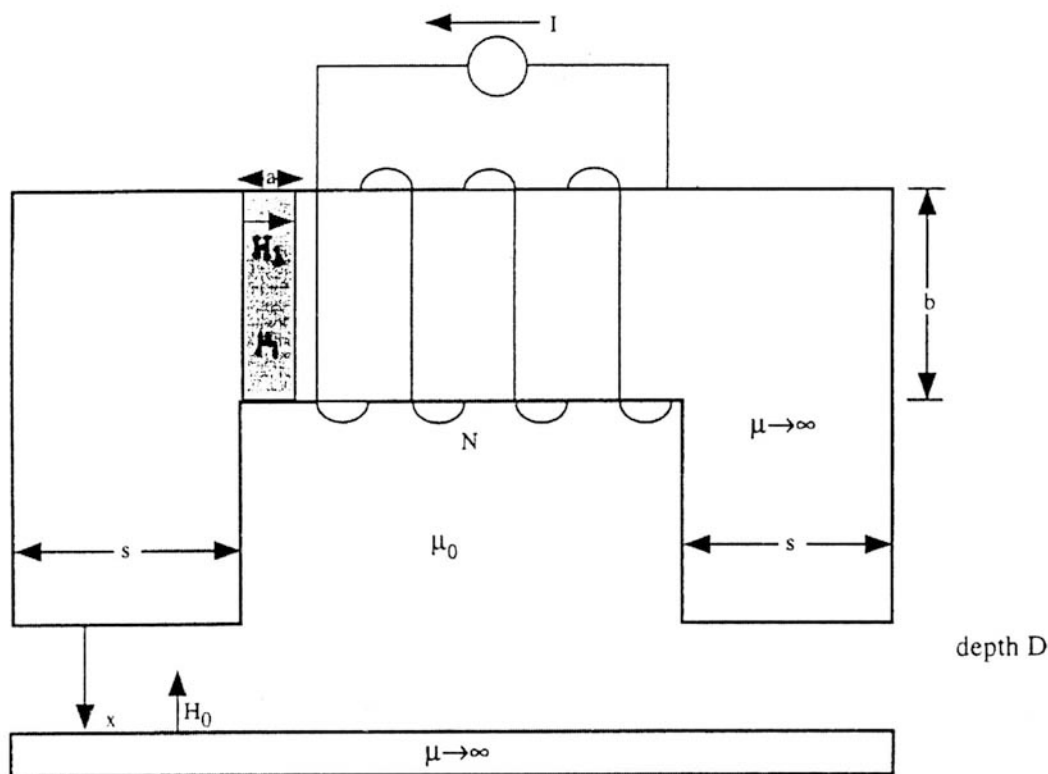


Figure 6: A magnetic circuit

In the magnetic circuit shown above, a current  $I$  flows in the  $N$  turn coil which is mounted on a material of infinite magnetic permeability ( $\mu \rightarrow \infty$ ) except for a thin gap of width  $a$  and height  $b$  which has finite magnetic permeability  $\mu_1$ . The lower plate has infinite magnetic permeability ( $\mu \rightarrow \infty$ ) and is at a distance  $x$  below the upper assembly. The magnetic materials are surrounded by free space with magnetic permeability  $\mu_0$ . The entire system has depth  $D$ .

(A) Neglecting fringing field effects, find the magnetic field  $H_0$  in the air gap and  $H_1$  in the thin section of the upper magnetic part.

**Solution:**

$$\begin{aligned}
 H_0 2x + H_1 a &= NI \\
 \mu_0 H_0 s D &= \mu_1 H_1 b D \\
 H_1 &= \frac{\mu_0 H_0 s}{\mu_1 b}
 \end{aligned}$$



$$H_0 \left[ 2x + \frac{\mu_0 sa}{\mu_1 b} \right] = NI \Rightarrow H_0 = \frac{NI\mu_1 b}{2\mu_1 bx + \mu_0 sa}$$

$$H_1 = \frac{NI\mu_0 s}{2\mu_1 bx + \mu_0 sa}$$

(B) Find the self-inductance of the  $N$  turn coil.

**Solution:**

$$L = \frac{N\Phi}{I}, \Phi = \mu_0 H_0 s D = \mu_1 H_1 b D$$

$$= \frac{N^2 \mu_0 \mu_1 b s D}{2\mu_1 bx + \mu_0 sa}$$

(C) Find the total magnetic energy stored in the system.

**Solution:**

$$W_M = \frac{1}{2} LI^2 = \frac{1}{2} \frac{(NI)^2 \mu_0 \mu_1 b s D}{2\mu_1 bx + \mu_0 sa}$$

Alternate Method:

$$W_M = \frac{1}{2} \mu_0 H_0^2 x s D + \frac{1}{2} \mu_1 H_1^2 a b D$$

$$= H_0^2 \left[ \mu_0 x s D + \frac{1}{2} \mu_1 a b D \left( \frac{\mu_0 s}{\mu_1 b} \right)^2 \right]$$

$$= \frac{H_0^2 D}{(\mu_1 b)^2} \left[ \mu_0 x s (\mu_1 b)^2 + \frac{1}{2} \mu_1 a b (\mu_0 s)^2 \right]$$

$$= \frac{H_0^2 D}{(\mu_1 b)^2} \mu_0 \mu_1 b s \left[ x \mu_1 b + \frac{1}{2} \mu_0 a s \right]$$

$$= \frac{(NI)^2 (\mu_1^{\cancel{1}} b^{\cancel{1}}) D}{(\mu_1 b)^2 (2\mu_1 bx + \mu_0 sa)^2} \mu_0 \mu_1 b s \left[ x \mu_1 b + \frac{1}{2} \mu_0 a s \right]$$

$$= \frac{\frac{1}{2} (NI)^2 \mu_1 \mu_0 b s D}{(2\mu_1 bx + \mu_0 sa)}$$

(D) Find the magnetic force on the moveable lower plate as a function of  $x$ , material properties,  $N$ ,  $I$ , and geometric dimensions.

**Solution:**

$$f_x = \frac{1}{2} I^2 \frac{dL}{dx}$$

$$= -\frac{1}{2} \frac{I^2 N^2 \mu_0 \mu_1 b s D}{(2\mu_1 bx + \mu_0 sa)^2} 2\mu_1 b$$

$$= -\frac{(NI)^2 \mu_0 \mu_1^2 b^2 s D}{(2\mu_1 bx + \mu_0 sa)^2}$$

## Problem 2

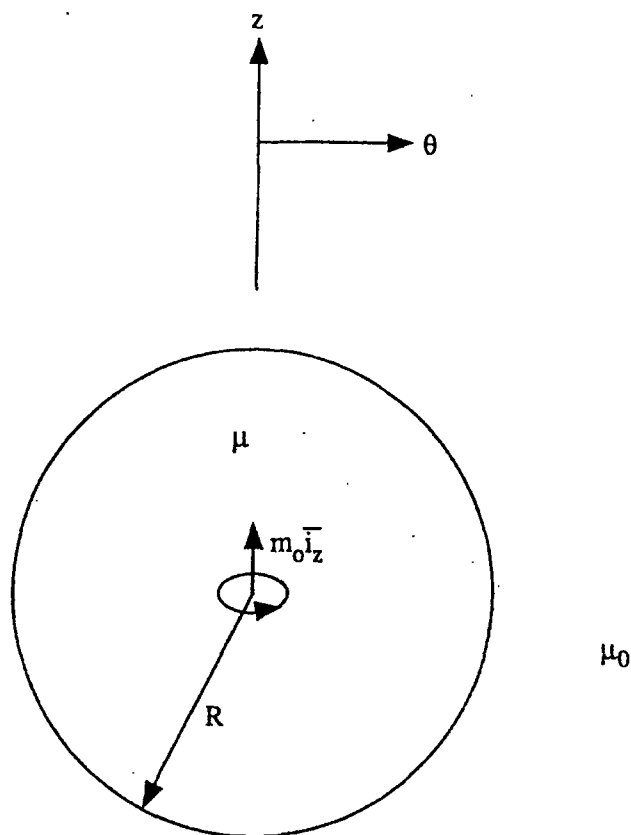


Figure 7: A sphere with a point magnetic dipole at its center

A point magnetic dipole with moment  $\vec{m} = m_0 \vec{i}_z$  is located at the center of a sphere of radius  $R$ . The sphere has finite magnetic permeability  $\mu$  and the sphere is surrounded by free space with magnetic permeability  $\mu_0$ . There is no free surface current on the  $r = R$  interface.

(A) What boundary conditions must be satisfied by the magnetic scalar potential and/or magnetic field at  $r = 0$ ,  $r = R$ , and  $r = \infty$ ?

**Solution:**

$$\chi_m(r=0) = \frac{m_0 \cos \theta}{4\pi r^2}, \chi_m(r \rightarrow \infty) = 0$$

$$H_\theta(r=R_-) = H_\theta(r=R_+), B_r(r=R_-) = B_r(r=R_+) \Rightarrow \mu H_r(r=R_-) = \mu_0 H_r(r=R_+)$$

(B) Find the magnetic field  $\vec{H}$  inside and outside the sphere.

**Solution:**

$$\chi_m = \begin{cases} \frac{m_0 \cos \theta}{4\pi r^2} + Ar \cos \theta & 0 < r < R \\ \frac{C}{r^2} \cos \theta & r > R \end{cases}$$

$$\vec{H} = -\nabla\chi_m = -\left[\frac{\partial\chi_m}{\partial r}\vec{i}_r + \frac{1}{r}\frac{\partial\chi_m}{\partial\theta}\vec{i}_\theta\right]$$

$$= \begin{cases} \frac{m_0}{4\pi r^3} (2 \cos \theta \vec{i}_r + \sin \theta \vec{i}_\theta) - A [\cos \theta \vec{i}_r - \sin \theta \vec{i}_\theta] & 0 < r < R \\ \frac{C}{r^3} (2 \cos \theta \vec{i}_r + \sin \theta \vec{i}_\theta) & r > R \end{cases}$$

$$H_\theta(r = R_-) = H_\theta(r = R_+) \Rightarrow \frac{C}{R^3} = \frac{m_0}{4\pi R^3} + A$$

$$\mu H_r(r = R_-) = \mu_0 H_r(r = R_+) \Rightarrow \mu \left[ \frac{2m_0}{4\pi R^3} - A \right] = \frac{2\mu_0 C}{R^3}$$

$$\begin{aligned} \frac{m_0}{4\pi R^3} + A &= \frac{C}{R^3} \\ \frac{2m_0}{4\pi R^3} - A &= \frac{2\mu_0}{\mu R^3} C \end{aligned}$$

$$\frac{C}{R^3} \left(1 + \frac{2\mu_0}{\mu}\right) = \frac{3m_0}{4\pi R^3} \Rightarrow \frac{C}{R^3} = \frac{3m_0}{4\pi R^3 \left(1 + \frac{2\mu_0}{\mu}\right)}$$

$$A = \frac{C}{R^3} - \frac{m_0}{4\pi R^3} = \frac{m_0}{4\pi R^3} \left( \frac{3}{1 + \frac{2\mu_0}{\mu}} - 1 \right) = \frac{m_0}{4\pi R^3} \left( \frac{2 \left(1 - \frac{\mu_0}{\mu}\right)}{\left(1 + 2\frac{\mu_0}{\mu}\right)} \right)$$

$$\vec{H} = \begin{cases} \frac{m_0}{4\pi r^3} [2 \cos \theta \vec{i}_r + \sin \theta \vec{i}_\theta] - \frac{m_0}{2\pi R^3} \frac{\left(1 - \frac{\mu_0}{\mu}\right)}{1 + \frac{2\mu_0}{\mu}} (\cos \theta \vec{i}_r - \sin \theta \vec{i}_\theta) & 0 < r < R \\ \frac{3m_0}{4\pi r^3} \frac{(2 \cos \theta \vec{i}_r + \sin \theta \vec{i}_\theta)}{\left(1 + \frac{2\mu_0}{\mu}\right)} & \end{cases}$$

(C) What is the effective magnetic dipole moment of the sphere seen by an observer for  $r > R$ ?

**Solution:**

$$\frac{m_{\text{eff}}}{4\pi} = C \Rightarrow m_{\text{eff}} = 4\pi \frac{3m_0}{4\pi \left(1 + \frac{2\mu_0}{\mu}\right)} \Rightarrow m_{\text{eff}} = \frac{3m_0}{1 + \frac{2\mu_0}{\mu}}$$

### Problem 3

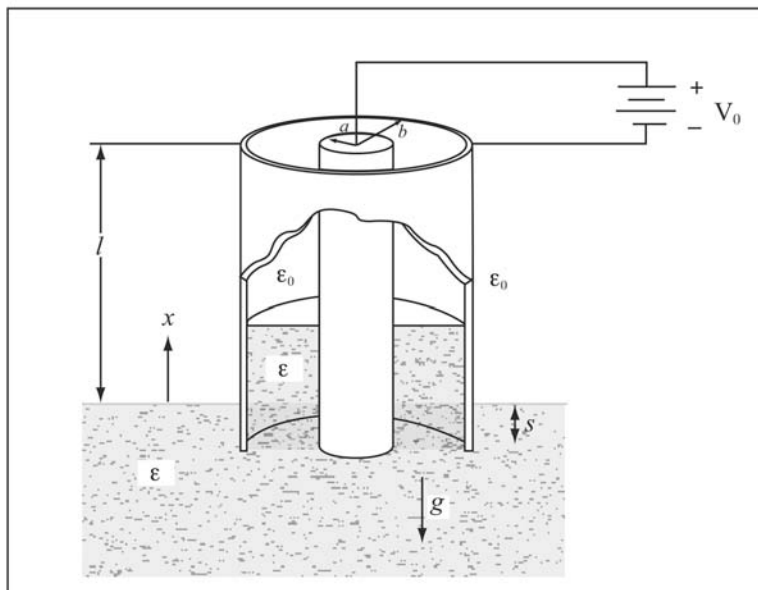


Figure 8: A coaxial cylindrical capacitor (Image by MIT OpenCourseWare.)

A coaxial cylindrical capacitor is dipped into a linearly polarizable fluid with dielectric permittivity  $\epsilon$  and mass density  $\rho_m$ . Gravity is directed downwards.

When voltage  $V_0$  is applied, the dielectric fluid is pulled into the coaxial capacitor to a height  $x$  above the fluid level outside the cylinders. If  $V_0 = 0$ , the fluid level within the cylinders is a distance  $s$  from the lower end of the cylinder. There is no free volume charge in the system.

**(A) Neglecting fringing field effects, what is the electric field, magnitude and direction, between the cylinders ( $a < r < b$ ) as a function of  $r$  in both the upper free space region and in the lower dielectric fluid?**

**Solution:**

$$\nabla \cdot \vec{E} = \frac{1}{r} \frac{d}{dr}(rE_r) = 0 \Rightarrow E_r = \frac{A}{r}$$

$$\int_a^b E_r dr = A \ln \frac{b}{a} = V_0 \Rightarrow A = \frac{V_0}{\ln \frac{b}{a}}$$

$$E_r = \frac{V_0}{r \ln \frac{b}{a}} \quad (\text{In both regions between cylinders})$$

Note that tangential  $E$  is continuous at the dielectric interface

**(B) What is the capacitance as a function of  $x$ ?**

**Solution:**

$$\text{In free space region: } D_r = \frac{\epsilon_0 V_0}{r \ln \frac{b}{a}}.$$

In dielectric fluid:  $D_r = \frac{\epsilon V_0}{r \ln \frac{b}{a}}$ .

Free surface charge on  $r=a$  surface:

$$\begin{aligned} q &= \epsilon_0 E_r \Big|_a 2\pi a(l-x-s) + \epsilon E_r \Big|_a 2\pi a(x+s) \\ &= E_r \Big|_a 2\pi a(\epsilon_0(l-x-s) + \epsilon(x+s)) \\ &= \frac{V_0}{\ln \frac{b}{a}} 2\pi \phi [\epsilon_0(l-x-s) + \epsilon(x+s)] \\ &= \frac{2\pi V_0}{\ln \frac{b}{a}} [\epsilon_0(l-x-s) + \epsilon(x+s)] \\ C &= \frac{q}{V_0} = \frac{2\pi [\epsilon_0(l-x-s) + \epsilon(x+s)]}{\ln \frac{b}{a}} \end{aligned}$$

(C) What is the total electric energy stored in the system?

**Solution:**

$$W_E = \frac{1}{2} C V_0^2 = \frac{\pi [\epsilon_0(l-x-s) + \epsilon(x+s)]}{\ln \frac{b}{a}} V_0^2$$

Alternate Method:

$$\begin{aligned} W_E &= \int_V \frac{1}{2} \epsilon E^2 dV = \int_{r=a}^b \int_{\Phi=0}^{2\pi} \int_{z=0}^l \frac{1}{2} \epsilon E^2 r dr d\Phi dz \\ W_E &= \int_{r=a}^b \frac{1}{2} E_r^2 r dr [\epsilon_0(l-x-s) + \epsilon(x+s)] 2\pi \\ &= \frac{2\pi [\epsilon_0(l-x-s) + \epsilon(x+s)] V_0^2}{2 \left[ \ln\left(\frac{b}{a}\right) \right]^2} \int_{r=a}^b \frac{dr}{r} \\ &= \frac{\pi [\epsilon_0(l-x-s) + \epsilon(x+s)] V_0^2}{\ln \frac{b}{a}} \end{aligned}$$

(D) How high will the dielectric fluid rise when a voltage  $V_0$  is applied?

**Solution:**

$$\begin{aligned} f_x &= \frac{1}{2} V_0^2 \frac{dC}{dx} = \frac{1}{2} \frac{V_0^2 2\pi [\epsilon - \epsilon_0]}{\ln \frac{b}{a}} = \rho_m g \pi (b^2 - a^2) x \\ x &= \frac{V_0^2 (\epsilon - \epsilon_0)}{\ln \frac{b}{a} \rho_m g (b^2 - a^2)} \end{aligned}$$

## Problem 4

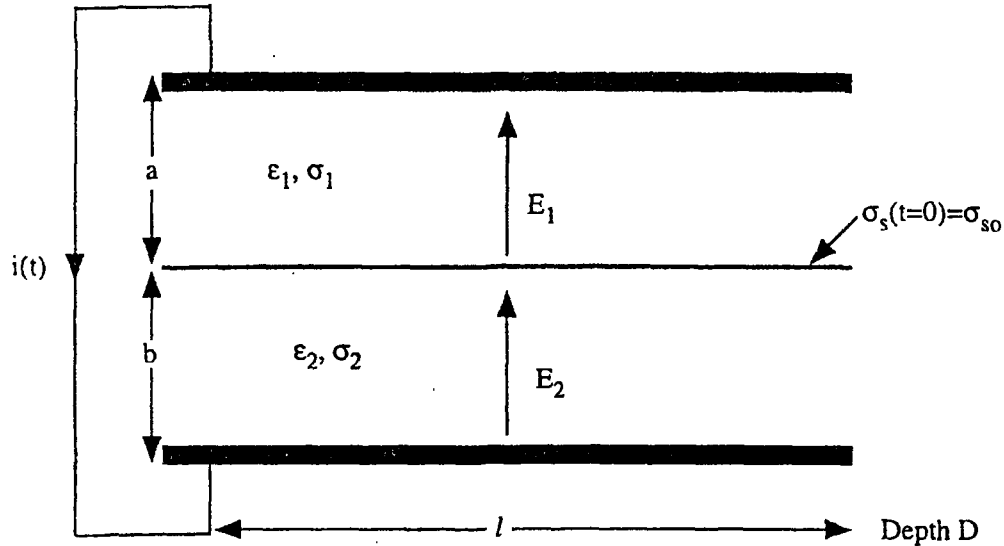


Figure 9: Two lossy dielectrics

Two lossy dielectrics with respective dielectric permittivities  $\epsilon_1$  and  $\epsilon_2$  and respective ohmic conductivities  $\sigma_1$  and  $\sigma_2$  are superposed within a short-circuited capacitor. At  $t = 0$  there is a free surface charge density of  $\sigma_{s0} \frac{C}{m^2}$  on the interface between the dielectrics. Neglect fringing field effects. The free volume charge density at time  $t = 0$  is zero in both dielectrics.

(A) Find the electric fields  $E_1(t = 0)$  and  $E_2(t = 0)$  in both lossy dielectrics at time  $t = 0$ .

**Solution:**

$$\rho_f(t) = 0 \text{ in both dielectrics} \Rightarrow \nabla \cdot \vec{E}_1 = \nabla \cdot \vec{E}_2 = 0 \Rightarrow E_1 = E_1(t), E_2 = E_2(t)$$

$$\int_{x=0}^{a+b} E dx = \int_{x=0}^b E_2 dx + \int_{x=b}^{a+b} E_1 dx = E_2 b + E_1 a = 0 \Rightarrow E_2 = -\frac{E_1 a}{b}$$

at  $t = 0$ :

$$\epsilon_1 E_1 - \epsilon_2 E_2 = \sigma_{s0}$$

$$E_1 \left[ \epsilon_1 + \frac{\epsilon_2 a}{b} \right] = \sigma_{s0}$$

$$E_1 = \frac{\sigma_{s0} b}{\epsilon_1 b + \sigma_2 a}, E_2 = -\frac{E_1 a}{b} = -\frac{\sigma_{s0} a}{\epsilon_1 b + \epsilon_2 a}$$

(B) Find the electric fields  $E_1(t)$  and  $E_2(t)$  in both lossy dielectrics as a function of time.

Solution:

$$\sigma_1 E_1 - \sigma_2 E_2 + \frac{d}{dt} [\epsilon_1 E_1 - \epsilon_2 E_2] = 0$$

$$E_1 \left[ \sigma_1 + \frac{\sigma_2 a}{b} \right] + \left( \epsilon_1 + \frac{\epsilon_2 a}{b} \right) \frac{dE_1}{dt} = 0$$

$$E_1 = E_1(t=0) e^{-\frac{t}{\tau}}; \tau = \frac{\epsilon_1 b + \epsilon_2 a}{\sigma_1 b + \sigma_2 a}$$

$$E_1 = \frac{\sigma_{s0} b}{\epsilon_1 b + \epsilon_2 a} e^{-\frac{t}{\tau}}$$

$$E_2 = -\frac{E_1 a}{b} = -\frac{\sigma_{s0} a}{\epsilon_1 b + \epsilon_2 a} e^{-\frac{t}{\tau}}$$

(C) Find the free surface charge density  $\sigma_s(t)$  on the interface as a function of time.

Solution:

$$\sigma_s(t) = \epsilon_1 E_1 - \epsilon_2 E_2 = E_1 \left( \epsilon_1 + \frac{\epsilon_2 a}{b} \right) = \sigma_{s0} e^{-\frac{t}{\tau}}$$

(D) Find the short circuit current  $i(t)$  that flows in the wire short-circuiting the two electrodes as a function of time.

Solution:

$$\begin{aligned} \frac{i(t)}{ld} &= \sigma_2 E_2 + \epsilon_2 \frac{dE_2}{dt} = \sigma_1 E_1 + \epsilon_1 \frac{dE_1}{dt} \\ &= \left( \sigma_1 - \frac{\epsilon_1}{\tau} \right) E_1 \\ &= \left( \sigma_1 - \frac{\epsilon_1}{(\epsilon_1 b + \sigma_2 a)} (\sigma_1 b + \sigma_2 a) \right) E_1 \\ &= \left( \frac{\sigma_1 \epsilon_1 b + \sigma_1 \epsilon_2 a - \epsilon_1 \sigma_1 b - \epsilon_1 \sigma_2 a}{\epsilon_1 b + \epsilon_2 a} \right) E_1 \\ &= \frac{a(\sigma_1 \epsilon_2 - \epsilon_1 \sigma_2)}{(\epsilon_1 b + \epsilon_2 a)^2} \sigma_{s0} b e^{-\frac{t}{\tau}} \\ i(t) &= \frac{ab\sigma_{s0}(\sigma_1 \epsilon_2 - \epsilon_1 \sigma_2)ld}{(\epsilon_1 b + \epsilon_2 a)^2} e^{-\frac{t}{\tau}} \end{aligned}$$

## Problem 5

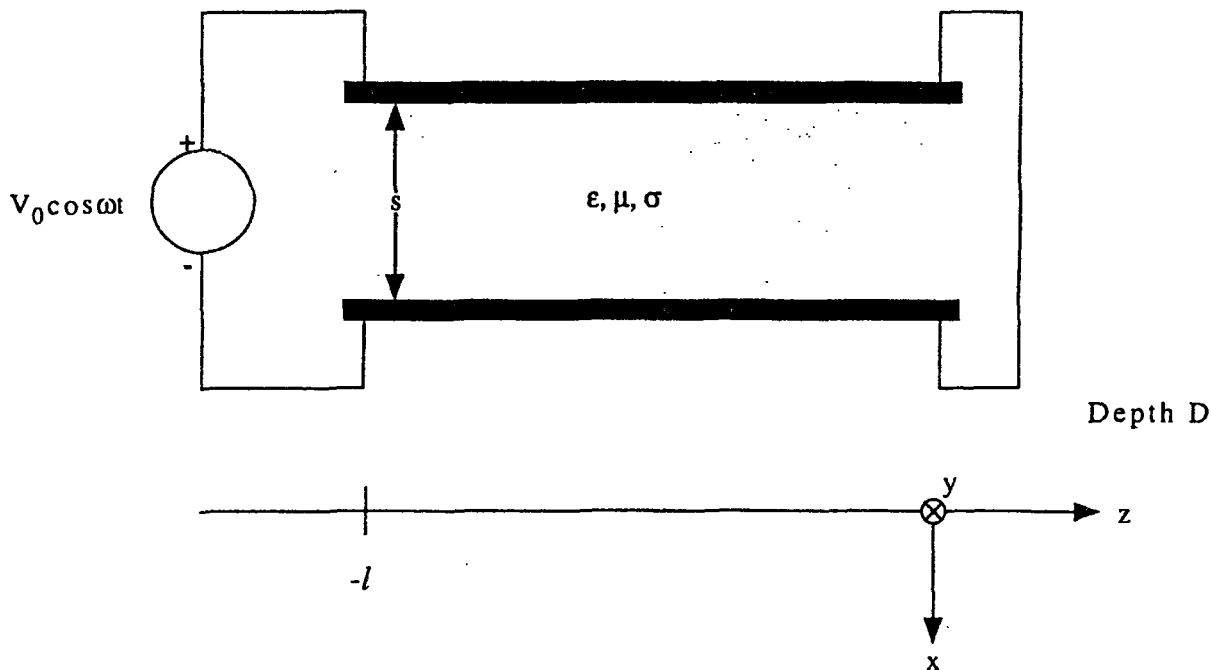


Figure 10: A lossy transmission line

A lossy transmission line is composed of perfectly conducting parallel plates enclosing a lossy medium with dielectric permittivity  $\epsilon$ , magnetic permeability  $\mu$ , and ohmic conductivity  $\sigma$ . The governing equations for the voltage  $v(z, t)$  and current  $i(z, t)$  along the transmission line are

$$\begin{aligned}\frac{\partial i}{\partial z} &= -C \frac{\partial v}{\partial t} - Gv \\ \frac{\partial v}{\partial z} &= -L \frac{\partial i}{\partial t}\end{aligned}$$

Where  $C$  is the capacitance per unit length,  $G$  is the conductance per unit length, and  $L$  is the inductance per unit length. The transmission line is short circuited at  $z = 0$  and is driven by a voltage source at  $z = -l$ ,  $v(z = -l, t) = V_0 \cos(\omega t)$ .

(A) What are  $C$ ,  $G$ , and  $L$  in terms of  $\epsilon$ ,  $\mu$ ,  $\sigma$ ,  $l$ ,  $D$  and  $s$ ?

**Solution:**

$$C = \frac{\epsilon d}{s}, L = \frac{\mu s}{d}, G = \frac{\sigma d}{s} \quad \left( RC = \frac{\epsilon}{\sigma} \Rightarrow G = \frac{1}{R} = \frac{\sigma C}{\epsilon} \right)$$



(B) In the sinusoidal steady state the voltage and current can be written in the form

$$v(z, t) = \text{Re} [\hat{v}(z)e^{j\omega t}]$$

$$i(z, t) = \text{Re} [\hat{i}(z)e^{j\omega t}]$$

Find  $\hat{v}(z)$  for this problem.

**Solution:**

$$\frac{d\hat{i}}{dz} = -(G + Cj\omega)\hat{v}$$

$$\frac{d\hat{v}}{dz} = -Lj\omega\hat{i} \Rightarrow \hat{i} = -\frac{1}{Lj\omega} \frac{d\hat{v}}{dz}$$

$$\frac{1}{Lj\omega} \frac{d^2\hat{v}}{dz^2} = +(G + Cj\omega)\hat{v}$$

$$\frac{d^2\hat{v}}{dz^2} = (GLj\omega - LC\omega^2)\hat{v}$$

$$\begin{aligned} \hat{v}(z) = Ae^{jpz} \Rightarrow -p^2 &= GLj\omega - LC\omega^2 \\ p &= \pm\sqrt{LC\omega^2 - GLj\omega} \\ p &= \pm p_0, p_0 = \sqrt{LC\omega^2 - GLj\omega} \end{aligned}$$

$$\hat{v}(z) = A_1 e^{jp_0 z} + A_2 e^{-jp_0 z}$$

$$\hat{v}(z=0) = 0 = A_1 + A_2$$

$$\hat{v}(z=-l) = V_0 = A_1 e^{-jp_0 l} + A_2 e^{jp_0 l} = A_1 (e^{-jp_0 l} - e^{jp_0 l}) = -2j A_1 \sin p_0 l$$

$$A_1 = -A_2 = -\frac{V_0}{2j \sin p_0 l}$$

$$\hat{v}(z) = -\frac{V_0}{2j \sin p_0 l} (e^{jp_0 z} - e^{-jp_0 z}) = -\frac{V_0 \sin p_0 z}{\sin p_0 l}$$

(C) Find  $\hat{i}(z)$  for this problem.

**Solution:**

$$\hat{i}(z) = -\frac{1}{Lj\omega} \frac{d\hat{v}}{dz} = -\frac{1}{Lj\omega} \left( -\frac{V_0 p_0 \cos p_0 z}{\sin p_0 l} \right) = \frac{V_0 p_0 \cos p_0 z}{Lj\omega \sin p_0 l}$$

(D) Now, assuming  $G = 0$  and neglecting fringing field effects, find the Poynting vector  $\vec{S} = \vec{E} \times \vec{H}$  as a function of time and position  $z$  everywhere along the transmission line for  $-l \leq z \leq 0$ .

**Solution:**

$$G = 0 \Rightarrow p_0 = \omega\sqrt{LC} \quad (\text{real}), v(z, t) = \text{Re}\hat{v}(z)e^{j\omega t} = \frac{V_0 \sin p_0 z}{\sin p_0 l} \cos \omega t$$

$$i(z, t) = \text{Re} \left[ + \frac{V_0 p_0 \cos p_0 z}{L j \omega \sin p_0 l} e^{j\omega t} \right] = \frac{V_0 p_0 \cos p_0 z}{L \omega \sin p_0 l} \sin \omega t = +V_0 \sqrt{\frac{C}{L}} \frac{\cos p_0 z}{\sin p_0 l} \sin \omega t$$

$$E_x = \frac{v(z, t)}{s}, H_y = \frac{i(z, t)}{d} \Rightarrow \vec{S} = \vec{E} \times \vec{H} = E_x H_y \hat{i}_z = \frac{v(z, t)i(z, t)}{sd} \hat{i}_z = \frac{-V_0^2 \sqrt{\frac{C}{L}} \sin p_0 z \cos p_0 z \sin \omega t \cos \omega t}{sd \sin^2 p_0 l} \hat{i}_z$$

## Solutions to Two More Problems

## Problem 1

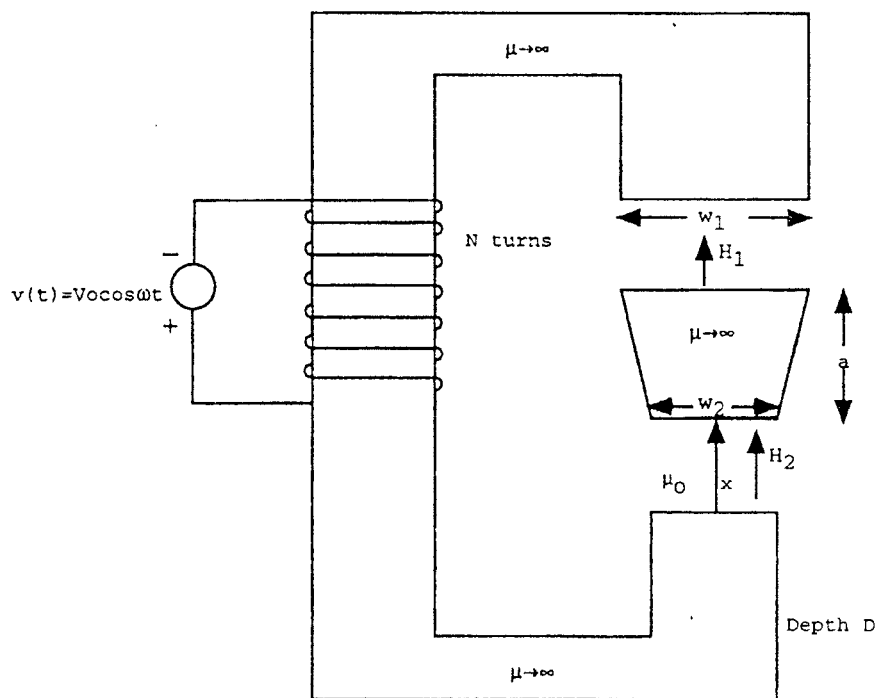


Figure 11: A magnetic circuit with a stationary yoke

A magnetic circuit has a stationary yoke with infinite magnetic permeability with a voltage source  $v(t) = V_0 \cos \omega t$  exciting an  $N$ -turn perfectly conducting coil. In the air gap of the magnetic yoke of height  $s$  there is an infinitely magnetically permeable tapered wedge of height  $a$  ( $a < s$ ) whose width decreases from  $w_1$  to  $w_2$ . The bottom surface of the wedge is a distance  $x$  above the lower surface of the magnetic yoke. The system has depth  $D$ . Assume that both air gaps are sufficiently small to neglect fringing fields.

(A) What is the total magnetic flux  $\lambda(t)$  linking the  $N$ -turn coil?

**Solution:**

$$v = V_0 \cos \omega t = \frac{d\lambda}{dt} \Rightarrow \lambda = \frac{V_0}{\omega} \sin \omega t$$

(B) What are the magnetic fields,  $\bar{H}_1(t)$  and  $\bar{H}_2(t)$ , in the air gaps in terms of  $\lambda(t)$ ,  $\mu_0$ , and geometric parameters?

**Solution:**

$$\mu_0 H_1 w_1 D = \mu_0 H_2 w_2 D = \frac{\lambda}{N}$$

$$H_1 = \frac{\lambda}{N \mu_0 w_1 D}$$

$$H_2 = \frac{\lambda}{N \mu_0 w_2 D}$$

(C) What is the self-inductance  $L(x)$  of the  $N$  turn coil as a function of the distance  $x$ ,  $\mu_0$ , and geometric parameters?

**Solution:**

$$\begin{aligned} \oint \vec{H} \cdot d\vec{l} &= Ni = H_1(s - a - x) + H_2 x \\ &= \frac{\lambda}{N \mu_0 D} \left( \frac{s - a - x}{w_1} + \frac{x}{w_2} \right) \end{aligned}$$

$$L(x) = \frac{\lambda}{i} = \frac{N^2 \mu_0 D}{\left[ \frac{s - a - x}{w_1} + \frac{x}{w_2} \right]}$$

(D) What is the  $x$ -directed force on the tapered wedge in terms of  $\lambda(t)$ ,  $\mu_0$ , and geometric parameters?

**Solution:**

$$\begin{aligned} f_x &= -\frac{1}{2} \lambda^2 \frac{d}{dx} \left[ \frac{1}{L(x)} \right] \\ &= -\frac{\lambda^2}{2} \frac{d}{dx} \frac{\left[ \frac{s - a - x}{w_1} + \frac{x}{w_2} \right]}{N^2 \mu_0 D} \\ &= -\frac{\lambda^2}{2 N^2 \mu_0 D} \left[ \frac{1}{w_2} - \frac{1}{w_1} \right] \end{aligned}$$

## Problem 5

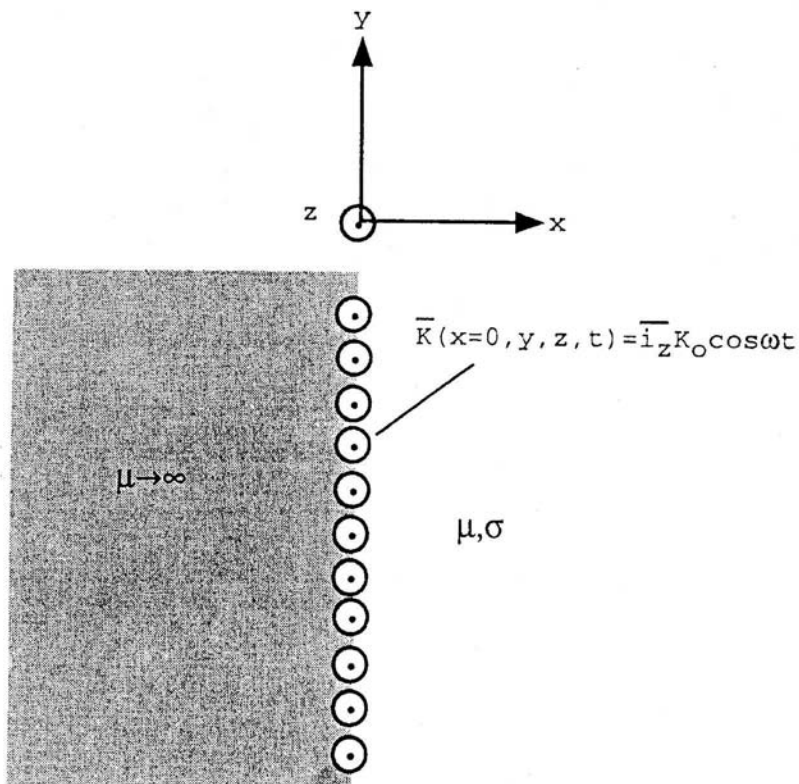


Figure 12: A current sheet of infinite extent

A  $z$  directed current sheet of infinite extent in the  $y$  and  $z$  directions is located at  $x = 0$  and varies with time as

$$\bar{K}(x = 0, y, z, t) = \bar{i}_z K_0 \cos \omega t$$

This current sheet is located at the interface separating a material of infinite magnetic permeability ( $\mu \rightarrow \infty$ ) for  $-\infty < x < 0$  and a material of finite magnetic permeability  $\mu$  and finite ohmic conductivity  $\sigma$  for  $0 < x < \infty$ . Note that because the current sheet has no variation with  $y$  or  $z$ , the magnetic field does not depend on the  $y$  or  $z$  coordinates.

(A) Find the magnitude and direction of the magnetic field  $\bar{H}(x, t)$  everywhere.

**Solution:**

$$H_y(x, t) = \text{Re} \hat{H}_y(x) e^{j\omega t}$$

$$\hat{H}_y(x = 0) = K_0 \Rightarrow \hat{H}_y(x) = K_0 e^{-\frac{(1+j)x}{\delta}}, \delta = \sqrt{\frac{2}{\omega \mu \sigma}}$$

$$\begin{aligned} H_y(x, t) &= \text{Re} K_0 e^{-\frac{x}{\delta}} e^{-\frac{jx}{\delta}} e^{j\omega t} \\ &= K_0 e^{-\frac{x}{\delta}} \cos\left(\omega t - \frac{x}{\delta}\right) \quad x > 0 \end{aligned}$$

$$H_y(x, t) = 0 \quad x < 0$$

(B) Find the volume current density  $\bar{J}(x, t)$  everywhere.

**Solution:**

$$\nabla \times \bar{H} = \bar{J} \Rightarrow \frac{\partial H_y}{\partial x} = J_z, J_z = 0, x < 0$$

$$J_z = \frac{\partial H_y}{\partial x} = \frac{K_0 e^{-\frac{x}{\delta}}}{\delta} \left[ -\cos\left(\omega t - \frac{x}{\delta}\right) + \sin\left(\omega t - \frac{x}{\delta}\right) \right]$$

(C) Find the power flow density,  $\bar{S} = \bar{E} \times \bar{H}$ , everywhere.

**Solution:**

$x > 0$ :

$$\begin{aligned} \bar{S} = \bar{E} \times \bar{H} &= \frac{\bar{J}}{\sigma} \times \bar{H} = \frac{J_z}{\sigma} H_y \bar{i}_z \times \bar{i}_y \\ &= -\frac{J_z}{\sigma} H_y \bar{i}_x \\ &= -\frac{K_0^2 e^{-\frac{2x}{\delta}}}{\delta \sigma} \left[ -\cos\left(\omega t - \frac{x}{\delta}\right) + \sin\left(\omega t - \frac{x}{\delta}\right) \right] \cos\left(\omega t - \frac{x}{\delta}\right) \\ &= \frac{K_0^2}{\sigma \delta} e^{-\frac{2x}{\delta}} \cos\left(\omega t - \frac{x}{\delta}\right) \left[ \cos\left(\omega t - \frac{x}{\delta}\right) - \sin\left(\omega t - \frac{x}{\delta}\right) \right] \end{aligned}$$

$$\bar{S} = 0, x < 0$$

## Final Exam 1995 Solutions

## Problem 1

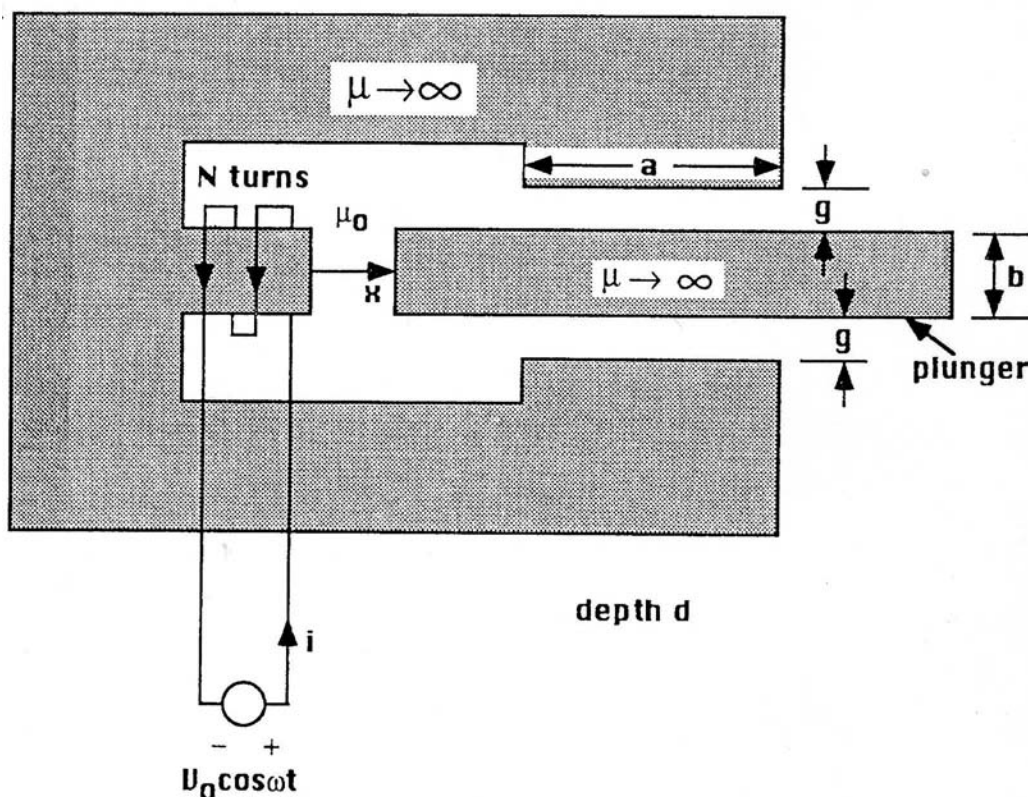


Figure 13: A magnetic circuit

The magnetic circuit shown above is modeled as being infinitely permeable except for the three thin air-gaps, where  $\mu = \mu_0$ . These thin gaps are narrow enough that fringing fields can be ignored. The  $N$  turn coil is driven by the voltage source  $v(t) = V_0 \cos \omega t$ .

(A) Determine the self-inductance  $L(x)$  of the  $N$  turn coil.

**Solution:**

$$H_g g + H_x x = Ni$$

$$\mu_0 H_x b d = 2\mu_0 H_g a d \Rightarrow H_g = \frac{H_x b}{2a}$$

$$H_x \left[ x + \frac{gb}{2a} \right] = Ni \Rightarrow H_x = \frac{Ni}{\left[ x + \frac{gb}{2a} \right]}$$

$$V_0 \cos \omega t = \frac{d\lambda}{dt} \Rightarrow \lambda = \frac{V_0}{\omega} \sin \omega t = N \mu_0 H_x b d = \frac{\mu_0 b d N^2 i}{\left[ x + \frac{gb}{2a} \right]}$$

$$L(x) = \frac{\lambda}{i} = \frac{\mu_0 b d N^2}{\left[x + \frac{gb}{2a}\right]}$$

(B) Find the total magnetic energy stored in the system as a function of time  $t$  in terms of  $V_0$ ,  $\omega$ , and given geometric and physical parameters.

**Solution:**

$$W_m = \frac{1}{2} \frac{\lambda^2}{L(x)} = \frac{1}{2} \frac{V_0^2 \sin^2 \omega t}{\omega^2 \mu_0 b d N^2} \left[ x + \frac{gb}{2a} \right]$$

(C) Determine the magnetic force acting on the movable plunger in the  $x$  direction as a function of time  $t$  in terms of  $V_0$ ,  $\omega$ , and given geometric and physical parameters.

**Solution:**

$$f = -\frac{1}{2} \lambda^2 \frac{d}{dx} \left[ \frac{1}{L(x)} \right] = -\frac{V_0^2 \sin^2 \omega t}{2\omega^2 \mu_0 b d N^2}$$



### Problem 5

A sphere of magnetic material having radius  $R$  is to be magnetized by placing it in a source of uniform magnetic field intensity. The bulk of the sphere has a constant magnetic permeability  $\mu$  with zero electrical conductivity,  $\sigma = 0$ . The magnetizable sphere is surrounded by a thin spherical shell of material with thickness  $\Delta \ll R$  having electrical conductivity  $\sigma$  and magnetic permeability  $\mu_0$ . The field source is switched on at  $t = 0$  so that  $\bar{H}_0(t) = H_0 u(t) \bar{i}_z$  where  $u(t)$  is the unit step function in time.

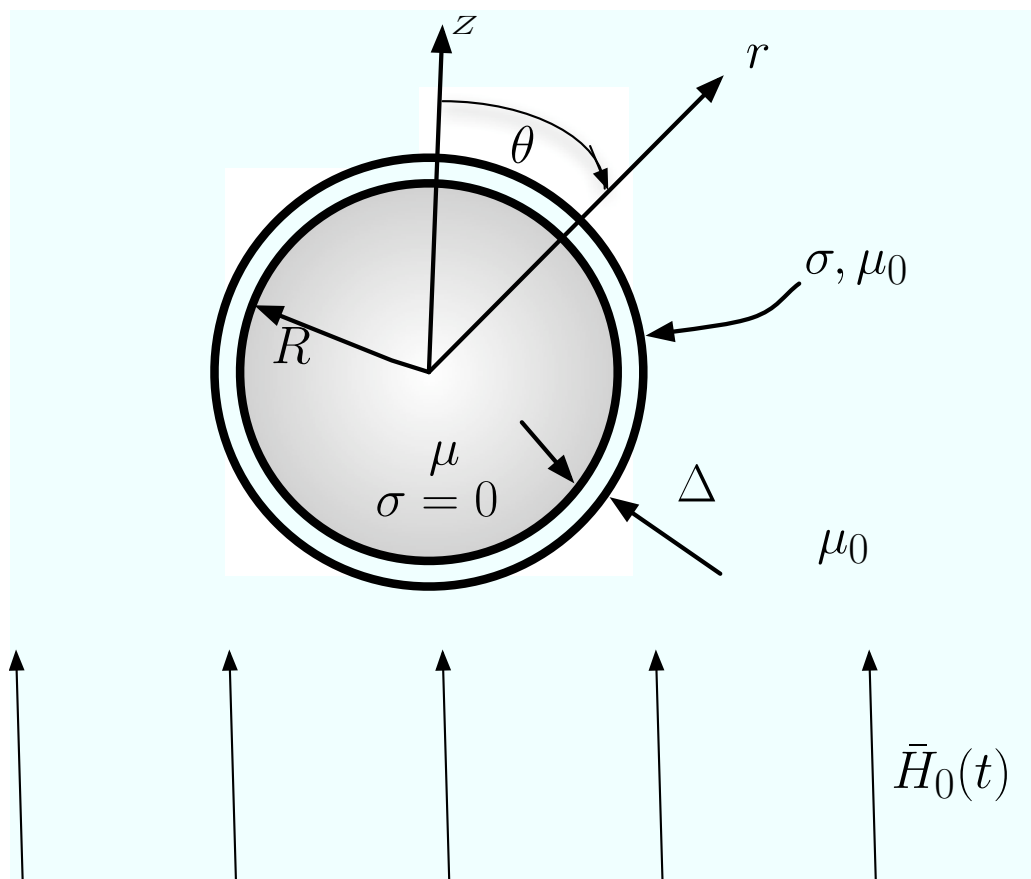


Figure 14: A sphere of magnetic material with a non-magnetic conducting coating (Image by MIT OpenCourseWare.)

(A) What is the magnetic field intensity  $\bar{H}$  inside the magnetizable sphere for  $r < R$  at  $t = 0^+$  and at  $t \rightarrow \infty$ ?

**Solution:**

$$\bar{H}(t = 0_+) = 0 \quad r < R$$

$$t \rightarrow \infty : \bar{H} = -\nabla\chi = -\left[ \frac{\partial\chi}{\partial r} \bar{i}_r + \frac{1}{r} \frac{\partial\chi}{\partial\theta} \bar{i}_\theta \right]$$

$$\nabla^2\chi = 0$$

$$\chi = \begin{cases} Ar \cos \theta & 0 < r < R \\ (Cr + \frac{D}{r^2}) \cos \theta & r > R \end{cases}$$

$$\lim_{r \rightarrow \infty} \bar{H} = H_0 \bar{i}_z = H_0 [\cos \theta \bar{i}_r - \sin \theta \bar{i}_\theta]$$

$$\lim_{r \rightarrow \infty} \chi = -H_0 z = -H_0 r \cos \theta$$

$$C = -H_0$$

$$\bar{H} = \begin{cases} -A [\cos \theta \bar{i}_r - \sin \theta \bar{i}_\theta] & 0 < r < R \\ - [(-H_0 - \frac{2D}{r^3}) \cos \theta \bar{i}_r - (-H_0 + \frac{D}{r^3}) \sin \theta \bar{i}_\theta] & r > R \end{cases}$$

$$H_\theta(r = R_-) = H_\theta(r = R_+) \Rightarrow A = -H_0 + \frac{D}{R^3}$$

$$\mu H_r(r = R_-) = \mu_0 H_r(r = R_+) \Rightarrow -\mu A = -\mu_0 \left( -H_0 - \frac{2D}{R^3} \right)$$

$$-\frac{\mu}{\mu_0} A = H_0 + \frac{2D}{R^3}$$

$$A = -H_0 + \frac{D}{R^3}$$

$$A = -\frac{3H_0}{2 + \frac{\mu}{\mu_0}}$$

$$\bar{H}(r < R, t \rightarrow \infty) = \frac{3H_0}{2 + \frac{\mu}{\mu_0}} \bar{i}_z$$

(B) The radial component of Faraday's law for this problem is:

$$\nabla \times \bar{E} = -\frac{\partial \bar{B}}{\partial t} \rightarrow \frac{1}{r \sin \theta} \frac{\partial}{\partial \theta} (\sin \theta E_\phi) = -\frac{\partial B_r}{\partial t}$$

Because  $\Delta \ll R$ , the current flow in the conducting spherical shell can be modeled as a surface current,  $K_\phi(r = R)$ . What is the approximate boundary condition at  $r = R$  relating the tangential ( $\theta$ ) component of  $\bar{H}$  on either side of the spherical shell to the perpendicular (radial) component of  $\bar{B}$ ?

**Solution:**

$$\nabla \times \bar{E} = -\frac{\partial \bar{B}}{\partial t} \Rightarrow \frac{1}{r \sin \theta} \frac{\partial}{\partial \theta} (\sin \theta E_\phi) = -\frac{\partial B_r}{\partial t}$$

In spherical shell:

$$J_\phi = \sigma E_\phi = \frac{K_\phi}{\Delta} = \frac{1}{\Delta} [H_\theta(r = R_+) - H_\theta(r = R_-)]$$

At  $r = R$ :

$$\frac{1}{\sigma \Delta R \sin \theta} \frac{\partial}{\partial \theta} [\sin \theta (H_\theta(r = R_+) - H_\theta(r = R_-))] = -\frac{\partial B_r}{\partial t}$$

(C) What is the approximate magnetic diffusion time  $\tau_m$  for this configuration?

**Solution:**

$$H_\theta(r = R_+) - H_\theta(r = R_-) = \left(-H_0 + \frac{D}{R^3}\right) \sin \theta - A \sin \theta$$

$$\mu H_r(r = R_-) = \mu_0 H_r(r = R_+) \Rightarrow -\mu A = -\mu_0 \left(-H_0 - \frac{2D}{R^3}\right) - \frac{1}{2} \left(\frac{\mu}{\mu_0} A + H_0\right) = \frac{D}{R^3}$$

$$H_\theta(r = R_+) - H_\theta(r = R_-) = \sin \theta \left[-H_0 - A + \frac{D}{R^3}\right] = \sin \theta \left[-H_0 - A - \frac{1}{2} \frac{\mu}{\mu_0} A - \frac{H_0}{2}\right]$$

$$\frac{1}{\sigma \Delta R \sin \theta} \frac{d}{d\theta} \left[ \sin^2 \theta \left(-\frac{3H_0}{2} - A \left(\frac{1}{2} \frac{\mu}{\mu_0} + 1\right)\right) \right] = \mu \frac{\partial A}{\partial t} \cos \theta$$

$$\frac{2}{\sigma \Delta R \mu} \left[-\frac{3H_0}{2} - A \left(\frac{1}{2} \frac{\mu}{\mu_0} + 1\right)\right] = \frac{\partial A}{\partial t} \Rightarrow \frac{dA}{dt} + \frac{2A}{\sigma \Delta R \mu} \left(\frac{1}{2} \frac{\mu}{\mu_0} + 1\right) = -\frac{3H_0}{\sigma \Delta R \mu}$$

$$\tau_m = \frac{\sigma \Delta R \mu}{\left(\frac{\mu}{\mu_0} + 2\right)}$$

## Final Exam 2000 Solutions

## Problem 1

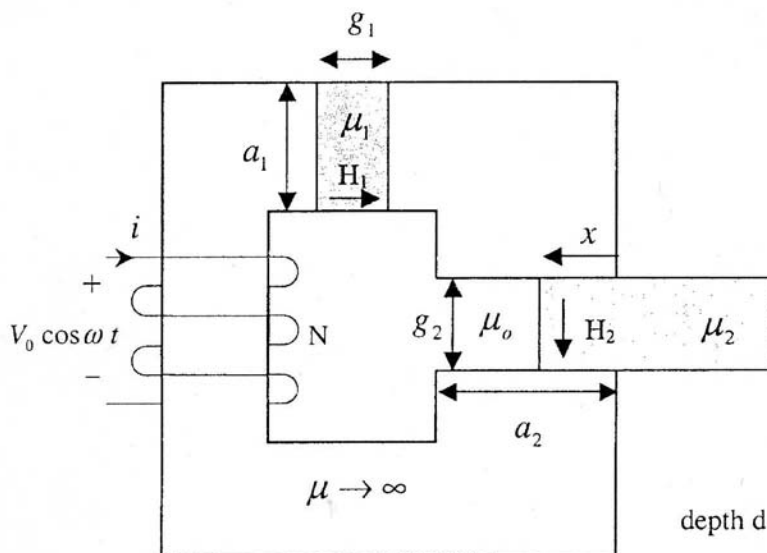


Figure 15: A magnetic circuit with a gap

The magnetic circuit shown above is modeled as being infinitely permeable except for the gap  $g_1$  of material with magnetic permeability  $\mu_1$ , and the free space gap partially filled with material with magnetic permeability  $\mu_2$ . The two gaps are sufficiently narrow that fringing fields are negligible. The  $N$  turn coil is driven by the voltage source  $v(t) = V_0 \cos \omega t$ .

(A) What is the magnetic flux  $\lambda$  through the  $N$  turn coil in terms of the terminal voltage?

**Solution:**

$$v = V_0 \cos \omega t = \frac{d\lambda}{dt} \Rightarrow \lambda = \frac{V_0}{\omega} \sin \omega t$$

(B) What are the magnetic fields  $H_1$  and  $H_2$  in the two gaps in terms of the magnetic flux,  $\lambda$ , magnetic permeabilities, and geometric factors?

**Solution:**

$$\Phi = \frac{\lambda}{N} = \mu_1 H_1 a_1 d = H_2 d (\mu_2 x + \mu_0 (a_2 - x))$$

$$H_1 = \frac{\lambda}{a} N \mu_1 a_1 d, H_2 = \frac{\lambda}{Nd [\mu_2 x + \mu_0 (a_2 - x)]}$$

(C) What is the coil current  $i$ ?

**Solution:**

$$H_1 g_1 + H_2 g_2 = Ni = \frac{\lambda g_1}{N\mu_1 a_1 d} + \frac{\lambda g_2}{Nd[\mu_2 x + \mu_0(a_2 - x)]}$$

$$i = \frac{\lambda}{N^2 d} \left[ \frac{g_1}{\mu_1 a_1} + \frac{g_2}{[\mu_2 x + \mu_0(a_2 - x)]} \right]$$

(D) What is the self-inductance  $L(x)$  of the  $N$  turn coil where  $x$  is the penetration distance of the material with magnetic permeability  $\mu_2$  into the free space gap?

**Solution:**

$$L(x) = \frac{\lambda}{i} = \frac{N^2 d}{\left[ \frac{g_1}{\mu_1 a_1} + \frac{g_2}{[\mu_2 x + \mu_0(a_2 - x)]} \right]}$$

(E) What is the magnetic stored energy?

**Solution:**

$$W_m = \frac{1}{2} L(x) i^2 = \frac{1}{2} \frac{\lambda^2}{L(x)}$$

(F) Determine the magnitude and direction of the magnetic force on the movable slab with magnetic permeability  $\mu_2$ .

**Solution:**

$$f_x = \frac{1}{2} i^2 \frac{dL}{dx} = -\frac{\lambda^2}{2} \frac{d}{dx} \left( \frac{1}{L(x)} \right)$$

$$\frac{1}{L(x)} = \frac{\left[ \frac{g_1}{\mu_1 a_1} + \frac{g_2}{[\mu_2 x + \mu_0(a_2 - x)]} \right]}{N^2 d}$$

$$f_x = -\frac{\lambda^2}{2N^2 d} g_2 \frac{-(\mu_2 - \mu_0)}{[\mu_2 x + \mu_0(a_2 - x)]^2} = \frac{\lambda^2 g_2}{2N^2 d} \frac{(\mu_2 - \mu_0)}{[\mu_2 x + \mu_0(a_2 - x)]^2}$$

## Problem 2

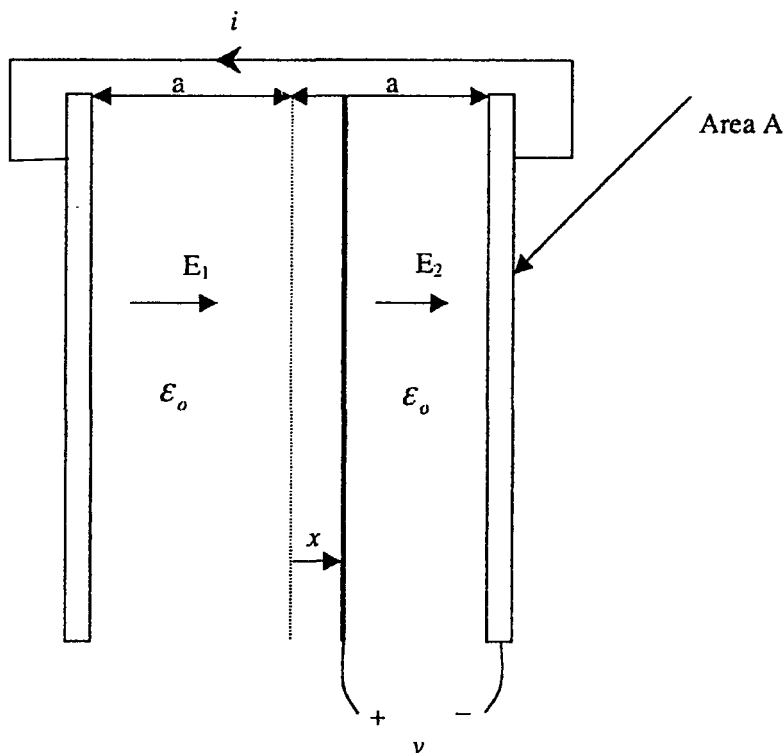


Figure 16: Short circuited parallel plate electrodes

Two parallel plate electrodes of area  $A$  in free space are a distance  $2a$  apart and are short circuited together. A third electrode at potential  $v$  with respect to the other two electrodes and with negligible thickness is placed at a distance  $x$  to the right of the midpoint position of the two short circuited electrodes.

(A) Find the electric fields  $E_1$  and  $E_2$  on either side of the middle electrode. Neglect fringing field effects.

**Solution:**

$$E_1 = -\frac{v}{a+x}, E_2 = \frac{v}{a-x}$$

(B) What is the total charge on the middle electrode?

**Solution:**

$$q_{\text{mid}} = \epsilon_0(E_2 - E_1)A = \epsilon_0 v A \left( \frac{1}{a-x} + \frac{1}{a+x} \right) = \frac{2\epsilon_0 v A a}{(a^2 - x^2)}$$

(C) What is the capacitance of the middle electrode with respect to the short circuited electrodes?

**Solution:**

$$C = \frac{q_{\text{mid}}}{v} = \frac{2\epsilon_0 a A}{a^2 - x^2}$$

(D) If the voltage  $v = v(t)$  and position  $x = x(t)$  are functions of time, what is the current  $i$  flowing in the short circuit?

**Solution:**

$$i = \epsilon_0 A \frac{dE_1}{dt} = -\epsilon_0 A \left( \frac{1}{a+x} \frac{dv}{dt} - \frac{v}{(a+x)^2} \frac{dx}{dt} \right) = -\frac{\epsilon_0 A}{(a+x)} \left( \frac{dv}{dt} - \frac{v}{(a+x)} \frac{dx}{dt} \right)$$

(E) What is the electric force on the middle electrode as a function of  $x, v, \epsilon_0$ , and geometric parameters  $a$  and  $A$ ?

**Solution:**

$$f_x = \frac{1}{2} v^2 \frac{dC}{dx} = \frac{1}{2} v^2 (2\epsilon_0 a A) \left( -\frac{1(-2x)}{(a^2 - x^2)^2} \right) = \frac{2\epsilon_0 a A x v^2}{(a^2 - x^2)^2}$$

## Problem 4

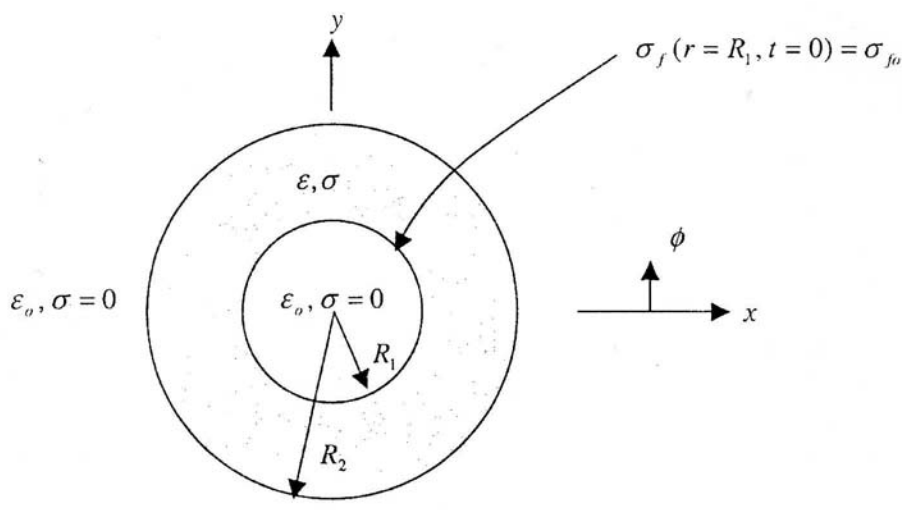


Figure 17: An infinitely long surface charged cylinder

An infinitely long cylinder with dielectric permittivity  $\epsilon$  and ohmic conductivity  $\sigma$  has outer radius  $R_2$  and free space hole of radius  $R_1$ . The cylinder is surrounded by free space for  $r > R_2$ . At time  $t = 0$  a uniform surface charge distribution is placed at  $r = R_1$  so that  $\sigma_f(r = R_1, t = 0) = \sigma_{f_0}$ . At time  $t = 0$  the free surface charge distribution at  $r = R_2$  is zero.

(A) What is the electric field in the regions  $r < R_1$ ,  $R_1 < r < R_2$  and  $r > R_2$  at time  $t = 0$ ?

**Solution:** At  $t = 0$ :

$$E_r = \begin{cases} 0 & r < R_1 \\ \frac{\sigma_{f_0} 2\pi R_1}{2\pi r \epsilon} & R_1 < r < R_2 \\ \frac{\sigma_{f_0} 2\pi R_1}{2\pi r \epsilon_0} & r > R_2 \end{cases}$$

(B) Find the electric field in the regions  $r < R_1$ ,  $R_1 < r < R_2$ , and  $r > R_2$  as a function of time.

**Solution:**

$$\sigma E_r(r = R_{1+}) + \epsilon \frac{\partial E_r(r = R_{1+})}{\partial t} = 0$$

$$E_r(r = R_{1+}, t) = E_r(r = R_{1+}, t = 0) e^{-\frac{t}{\tau}}; \tau = \frac{\epsilon}{\sigma}$$

$$\begin{aligned} \sigma_f(r = R_1) &= \epsilon E_r(r = R_{1+}, t) = \epsilon E_r(r = R_{1+}, t = 0) e^{-\frac{t}{\tau}} \\ &= \sigma_{f_0} e^{-\frac{t}{\tau}} \end{aligned}$$

$$E_r(r, t) = \begin{cases} 0 & r < R_1 \\ \frac{\sigma_{f_0} R_1}{\epsilon r} e^{-\frac{t}{\tau}} & R_1 < r < R_2 \\ \frac{\sigma_{f_0} R_1}{\epsilon_0 r} & r > R_2 \end{cases}$$



(C) Find the free surface charge distributions as a function of time at  $r = R_1$  and  $r = R_2$ .

**Solution:**

$$\begin{aligned}\sigma_f(r = R_1, t) &= \sigma_{f_0} e^{-\frac{t}{\tau}} \\ -\sigma E_r(r = R_{2-}, t) + \frac{\partial \sigma_f(r = R_2, t)}{\partial t} &= 0 \\ \frac{\partial \sigma_f(r = R_2, t)}{\partial t} = +\sigma E_r(r = R_{2-}, t) &= +\frac{\sigma}{\epsilon} \frac{\sigma_{f_0}}{R_2} R_1 e^{-\frac{t}{\tau}} \\ \sigma_f(r = R_2, t) &= +\frac{\sigma}{\epsilon} \frac{\sigma_{f_0} R_1}{R_2} (-\tau) e^{\frac{t}{\tau}} + C \\ &= \cancel{\sigma} \frac{\sigma_{f_0} R_1}{\cancel{\epsilon} R_2} \left( -\frac{\cancel{\tau}}{\cancel{\sigma}} \right) e^{-\frac{t}{\tau}} + C \\ &= \frac{-\sigma_{f_0} R_1}{R_2} e^{-\frac{t}{\tau}} + C \\ \sigma_f(r = R_2, t = 0) = 0 &= \frac{-\sigma_{f_0} R_1}{R_2} + C = 0 \Rightarrow C = \frac{\sigma_{f_0} R_1}{R_2} \\ \sigma_f(r = R_2, t) &= \frac{\sigma_{f_0} R_1}{R_2} \left( 1 - e^{-t/\tau} \right)\end{aligned}$$

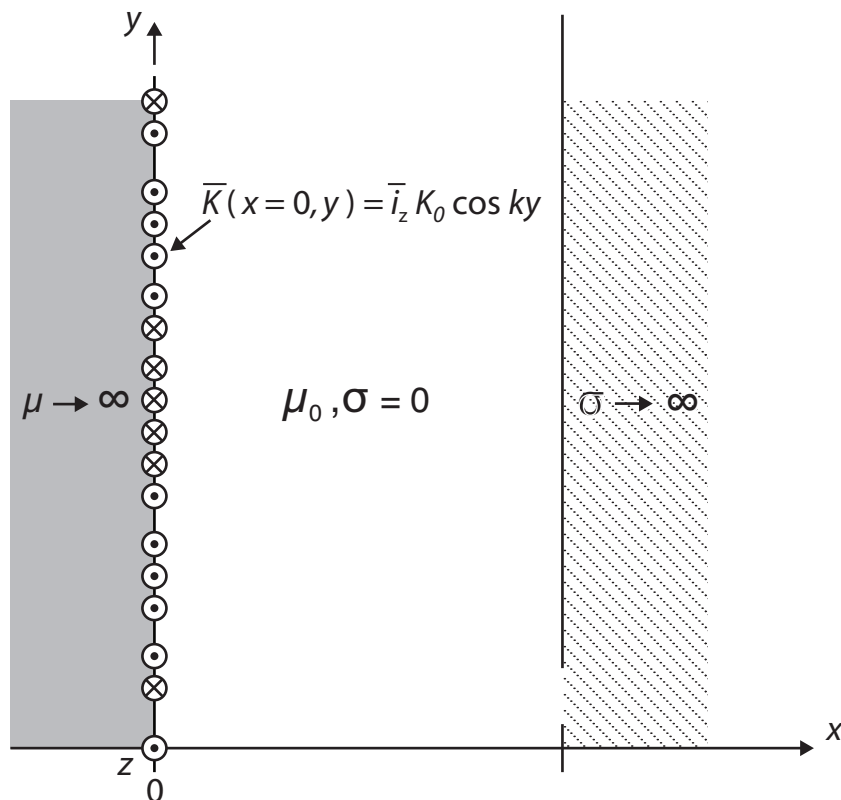
Another Way:

$$\begin{aligned}\sigma_f(r = R_2, t) &= \epsilon_0 E_r(r = R_{2+}, t) - \epsilon E_r(r = R_{2-}, t) \\ &= \frac{\sigma_{f_0} R_1}{R_2} - \frac{\sigma_{f_0} R_1}{R_2} e^{-\frac{t}{\tau}} \\ &= \frac{\sigma_{f_0} R_1}{R_2} \left( 1 - e^{-\frac{t}{\tau}} \right)\end{aligned}$$

Another way:

$$\begin{aligned}\sigma_f(r = R_1, t) 2\pi R_1 + \sigma_f(r = R_2, t) 2\pi R_2 &= \sigma_{f_0} (2\pi R_1) \\ \sigma_f(r = R_2, t) &= \frac{\sigma_{f_0} R_1}{R_2} - \sigma_f(r = R_1, t) \frac{R_1}{R_2} \\ &= \frac{\sigma_{f_0} R_1}{R_2} \left( 1 - e^{-\frac{t}{\tau}} \right)\end{aligned}$$

## Problem 5

Figure 18: A surface current sheet at  $x = 0$  (Image by MIT OpenCourseWare.)

A  $z$  directed surface current sheet of infinite extent in the  $y$  and  $z$  directions is located at  $x = 0$  and varies with coordinate  $y$  as  $\vec{K}(x = 0, y) = \hat{i}_z K_0 \cos ky$ . This current sheet is located at the  $x = 0$  interface separating a material of infinite magnetic permeability ( $\mu \rightarrow \infty$ ) for  $x < 0$  and free space for  $0 < x < s$ . At  $x = s$  there is another material of infinite extent for  $x > s$  with infinite ohmic conductivity ( $\sigma \rightarrow \infty$ ). There is no variation with the  $z$  coordinates and free space for  $0 < x < s$  is perfectly insulating ( $\sigma = 0$ ).

(A) What are the boundary conditions on the magnetic field  $\vec{H}(x, y)$  at  $x = 0$  and  $x = s$ ?

**Solution:**

$$H_y(x = 0_+) = K_0 \cos ky$$

$$H_x(x = s_-) = 0$$

(B) Find the magnetic field  $\vec{H}(x, y)$  everywhere.

**Solution:**

$$\chi(x, y) = \sin ky(Ae^{-kx} + Ce^{+kx}) \quad 0 < x < s$$

$$\bar{H} = -\nabla\chi = \begin{cases} 0 & x < 0 \\ 0 & x > s \\ -[-kAe^{-kx} + kCe^{+kx}] \sin ky \bar{i}_x - k \cos ky [Ae^{kx} + Ce^{+kx}] \bar{i}_y & 0 < x < s \end{cases}$$

$$H_x(x = s_-) = 0 \Rightarrow -kAe^{-ks} + kCe^{ks} = 0$$

$$H_y(x = 0_+) = K_0 \cos ky = -k \cos ky (A + C)$$

$$A + C = -\frac{K_0}{k}$$

$$A = Ce^{2ks} \Rightarrow C(1 + e^{2ks}) = -\frac{K_0}{k}$$

$$C = -\frac{\frac{K_0}{k}}{(1 + e^{2ks})}$$

$$A = -\frac{K_0 e^{2ks}}{k(1 + e^{2ks})}$$

$$\bar{H} = \begin{cases} 0 & x < 0 \\ 0 & x > s \\ -\frac{K_0}{(1 + e^{2ks})} [\sin(ky) (e^{-kx} e^{2ks} - e^{kx}) \bar{i}_x - \cos(ky) (e^{-kx} e^{2ks} + e^{kx}) \bar{i}_y] & 0 < x < s \end{cases}$$

$$0 < x < s$$

$$\begin{aligned} \bar{H} &= -\frac{2K_0 e^{ks}}{(1 + e^{2ks})} [\sin(ky) (-\sinh(k(x-s))) \bar{i}_x - \cos(ky) \cosh(k(x-s)) \bar{i}_y] \\ &= \frac{K_0}{\cosh(ks)} [\sin(ky) (\sinh(k(x-s))) \bar{i}_x + \cos(ky) \cosh(k(x-s)) \bar{i}_y] \end{aligned}$$

Check:

$$H_x(x = s) = 0$$

$$H_y(x = 0) = K_0 \cos(ky)$$

(C) What is the surface current on the  $x = s$  surface?

**Solution:**

$$K_z(x = s) = -H_y(x = s) = \frac{K_0 \cos(ky)}{\cosh(ks)}$$