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Lecture 8 - Transverse Electromagnetic Waves

I. Maxwell's Equations for Linear Media

$$\begin{aligned}\nabla \times \bar{E} &= -\mu \frac{\partial \bar{H}}{\partial t} \\ \nabla \times \bar{H} &= \bar{J} + \epsilon \frac{\partial \bar{E}}{\partial t} \\ \nabla \cdot \bar{E} &= \rho_f / \epsilon \\ \nabla \cdot \bar{H} &= 0\end{aligned}$$

II. Poynting's Theorem

A. Power Flow, Electromagnetic Energy

$$\begin{aligned}\nabla \cdot (\bar{E} \times \bar{H}) &= \bar{H} \cdot (\nabla \times \bar{E}) - \bar{E} \cdot (\nabla \times \bar{H}) \\ &= -\mu \bar{H} \cdot \frac{\partial \bar{H}}{\partial t} - \bar{E} \cdot \left[\bar{J} + \epsilon \frac{\partial \bar{E}}{\partial t} \right] \\ &= -\frac{\mu}{2} \frac{\partial}{\partial t} (|\bar{H}|^2) - \frac{\epsilon}{2} \frac{\partial}{\partial t} (|\bar{E}|^2) - \bar{E} \cdot \bar{J} \\ \nabla \cdot (\bar{E} \times \bar{H}) + \frac{\partial}{\partial t} \left[\frac{1}{2} \epsilon |\bar{E}|^2 + \frac{1}{2} \mu |\bar{H}|^2 \right] &= -\bar{E} \cdot \bar{J} \\ \int_V \nabla \cdot (\bar{E} \times \bar{H}) dV &= \oint_S \bar{E} \times \bar{H} \cdot d\bar{S} \\ \oint_S \bar{E} \times \bar{H} \cdot d\bar{a} + \frac{d}{dt} \int_V \left(\frac{1}{2} \epsilon |\bar{E}|^2 + \frac{1}{2} \mu |\bar{H}|^2 \right) dV &= - \int_V \bar{E} \cdot \bar{J} dV\end{aligned}$$

$$\bar{S} = \bar{E} \times \bar{H}$$

Poynting Vector (watts / m²)

$$W = \int_V \left(\frac{1}{2} \epsilon |\bar{E}|^2 + \frac{1}{2} \mu |\bar{H}|^2 \right) dV$$

Electromagnetic Stored Energy

$$P_d = \int_V \bar{E} \cdot \bar{J} dV$$

Power dissipated if $\bar{J} \cdot \bar{E} > 0$

$$e.g., \bar{J} = \sigma \bar{E} \Rightarrow \bar{J} \cdot \bar{E} = \sigma |\bar{E}|^2$$

Power source if $\bar{J} \cdot \bar{E} < 0$

$$P_{\text{out}} = \oint_S \bar{E} \times \bar{H} \cdot d\bar{a} = \oint_S \bar{S} \cdot d\bar{a}$$

$$P_{\text{out}} + \frac{dW}{dt} = -P_d$$

$$w_e = \frac{1}{2} \epsilon |\bar{E}|^2$$

Electric energy density in Joules/m³

$$w_m = \frac{1}{2} \mu |\bar{H}|^2$$

Magnetic energy density in Joules/m³

B. Power in Electric Circuits

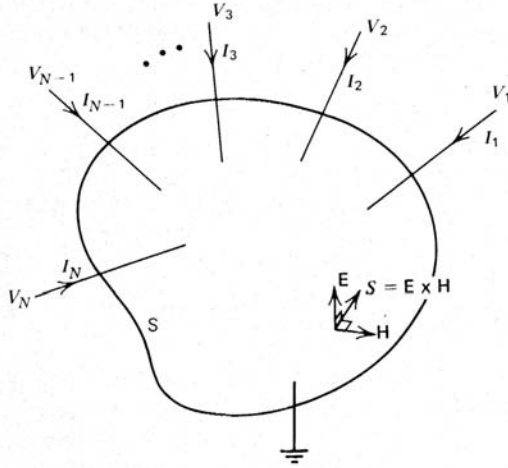


Figure 7-2 The circuit power into an N terminal network $\sum_{k=1}^N V_k I_k$ equals the electromagnetic power flow into the surface surrounding the network, $-\oint_S \mathbf{E} \times \mathbf{H} \cdot d\mathbf{S}$.

From *Electromagnetic Field Theory: A Problem Solving Approach*, by Markus Zahn, 1987. Used with permission.

Outside circuit elements

$$\oint_C \bar{\mathbf{E}} \cdot d\bar{\mathbf{l}} \approx 0, \quad \nabla \times \bar{\mathbf{E}} = 0 \Rightarrow \bar{\mathbf{E}} = -\nabla\Phi \quad (\text{Kirchoff's Voltage Law})$$

$$\sum_k v_k = 0$$

$$\nabla \times \bar{\mathbf{H}} = \bar{\mathbf{J}} \Rightarrow \nabla \cdot \bar{\mathbf{J}} = 0, \quad \oint_S \bar{\mathbf{J}} \cdot d\bar{\mathbf{S}} = 0 \quad (\text{Kirchoff's current law})$$

$$\sum_k i_k = 0$$

$$P_{\text{in}} = -\oint_S \bar{\mathbf{E}} \times \bar{\mathbf{H}} \cdot d\bar{\mathbf{S}}$$

$$= -\int_V \nabla \cdot (\bar{\mathbf{E}} \times \bar{\mathbf{H}}) dV$$

$$\nabla \cdot (\bar{\mathbf{E}} \times \bar{\mathbf{H}}) = \bar{\mathbf{H}} \cdot \nabla \times \bar{\mathbf{E}} - \bar{\mathbf{E}} \cdot (\nabla \times \bar{\mathbf{H}}) = -\bar{\mathbf{E}} \cdot \bar{\mathbf{J}} = \nabla\Phi \cdot \bar{\mathbf{J}}$$

$$\nabla \cdot (\bar{\mathbf{J}}\Phi) = \Phi \nabla \cdot \bar{\mathbf{J}} + \bar{\mathbf{J}} \cdot (\nabla\Phi)$$

$$\nabla \cdot (\bar{\mathbf{E}} \times \bar{\mathbf{H}}) = \bar{\mathbf{J}} \cdot (\nabla\Phi) = \nabla \cdot (\Phi\bar{\mathbf{J}})$$

$$P_{\text{in}} = -\int_V \nabla \cdot (\bar{\mathbf{E}} \times \bar{\mathbf{H}}) dV = -\int_V \nabla \cdot (\bar{\mathbf{J}}\Phi) dV = -\oint_S \bar{\mathbf{J}}\Phi \cdot d\bar{\mathbf{S}}$$

On S , Φ = voltages on each wire, $\bar{\mathbf{J}}$ is non-zero only on wires.

$$P_{\text{in}} = -\oint_S \bar{\mathbf{J}}\Phi \cdot d\bar{\mathbf{S}} = -\sum_{k=1}^N v_k \underbrace{\oint_S \bar{\mathbf{J}} \cdot d\bar{\mathbf{S}}}_{-i_k} = \sum_{k=1}^N v_k i_k$$

C. Complex Poynting's Theorem (Sinusoidal Steady State, $e^{j\omega t}$)

$$\bar{\mathbf{E}}(\bar{\mathbf{r}}, t) = \text{Re} \left[\hat{\mathbf{E}}(\bar{\mathbf{r}}) e^{j\omega t} \right] = \frac{1}{2} \left[\hat{\mathbf{E}}(\bar{\mathbf{r}}) e^{j\omega t} + \hat{\mathbf{E}}^*(\bar{\mathbf{r}}) e^{-j\omega t} \right]$$

$$\bar{\mathbf{H}}(\bar{\mathbf{r}}, t) = \text{Re} \left[\hat{\mathbf{H}}(\bar{\mathbf{r}}) e^{j\omega t} \right] = \frac{1}{2} \left[\hat{\mathbf{H}}(\bar{\mathbf{r}}) e^{j\omega t} + \hat{\mathbf{H}}^*(\bar{\mathbf{r}}) e^{-j\omega t} \right]$$

The real part of a complex number is one-half of the sum of the number and its complex conjugate

Maxwell's Equations in Sinusoidal Steady State

$$\begin{aligned}
 \nabla \times \hat{\mathbf{E}}(\bar{r}) &= -j\omega\mu\hat{\mathbf{H}}(\bar{r}) \\
 \nabla \times \hat{\mathbf{H}}(\bar{r}) &= \bar{\mathbf{J}}(\bar{r}) + j\omega\epsilon\hat{\mathbf{E}}(\bar{r}) \\
 \nabla \cdot \hat{\mathbf{E}}(\bar{r}) &= \hat{\rho}_f(\bar{r})/\epsilon \\
 \nabla \cdot \hat{\mathbf{H}}(\bar{r}) &= 0
 \end{aligned}$$

$$\begin{aligned}
 \bar{\mathbf{S}}(\bar{r}, t) &= \bar{\mathbf{E}}(\bar{r}, t) \times \bar{\mathbf{H}}(\bar{r}, t) \\
 &= \frac{1}{4} \left[\hat{\mathbf{E}}(\bar{r})e^{j\omega t} + \hat{\mathbf{E}}^*(\bar{r})e^{-j\omega t} \right] \times \left[\hat{\mathbf{H}}(\bar{r})e^{j\omega t} + \hat{\mathbf{H}}^*(\bar{r})e^{-j\omega t} \right] \\
 &= \frac{1}{4} \left[\hat{\mathbf{E}}(\bar{r}) \times \hat{\mathbf{H}}(\bar{r})e^{2j\omega t} + \hat{\mathbf{E}}^*(\bar{r}) \times \hat{\mathbf{H}}(\bar{r}) + \hat{\mathbf{E}}(\bar{r}) \times \hat{\mathbf{H}}^*(\bar{r}) + \hat{\mathbf{E}}^*(\bar{r}) \times \hat{\mathbf{H}}^*(\bar{r})e^{-2j\omega t} \right] \\
 \langle \bar{\mathbf{S}} \rangle &= \frac{1}{4} \left[\hat{\mathbf{E}}^*(\bar{r}) \times \hat{\mathbf{H}}(\bar{r}) + \hat{\mathbf{E}}(\bar{r}) \times \hat{\mathbf{H}}^*(\bar{r}) \right] \\
 &= \frac{1}{2} \text{Re} \left[\hat{\mathbf{E}}(\bar{r}) \times \hat{\mathbf{H}}^*(\bar{r}) \right] = \frac{1}{2} \text{Re} \left[\hat{\mathbf{E}}^*(\bar{r}) \times \hat{\mathbf{H}}(\bar{r}) \right]
 \end{aligned}$$

(A complex number plus its complex conjugate is twice the real part of that number.)

$$\begin{aligned}
 \hat{\mathbf{S}} &= \frac{1}{2} \hat{\mathbf{E}}(\bar{r}) \times \hat{\mathbf{H}}(\bar{r})^* \\
 \nabla \cdot \hat{\mathbf{S}} &= \nabla \cdot \left[\frac{1}{2} \hat{\mathbf{E}}(\bar{r}) \times \hat{\mathbf{H}}^*(\bar{r}) \right] = \frac{1}{2} \left[\hat{\mathbf{H}}^*(\bar{r}) \cdot \nabla \times \hat{\mathbf{E}}(\bar{r}) - \hat{\mathbf{E}}(\bar{r}) \cdot \nabla \times \hat{\mathbf{H}}^*(\bar{r}) \right] \\
 &= \frac{1}{2} \left[\hat{\mathbf{H}}^*(\bar{r}) \cdot \left(-j\omega\mu\hat{\mathbf{H}}(\bar{r}) \right) - \hat{\mathbf{E}}(\bar{r}) \cdot \left(\hat{\mathbf{J}}^*(\bar{r}) - j\omega\epsilon\hat{\mathbf{E}}^*(\bar{r}) \right) \right] \\
 &= \frac{1}{2} \left[-j\omega\mu|\hat{\mathbf{H}}(\bar{r})|^2 + j\omega\epsilon|\hat{\mathbf{E}}(\bar{r})|^2 \right] - \frac{1}{2} \hat{\mathbf{E}}(\bar{r}) \cdot \hat{\mathbf{J}}^*(\bar{r})
 \end{aligned}$$

$$\langle w_m \rangle = \frac{1}{4} \mu |\hat{\mathbf{H}}(\bar{r})|^2, \quad \langle w_e \rangle = \frac{1}{4} \epsilon |\hat{\mathbf{E}}(\bar{r})|^2$$

$$\hat{P}_d = \frac{1}{2} \hat{\mathbf{E}}(\bar{r}) \cdot \hat{\mathbf{J}}^*(\bar{r})$$

$$\nabla \cdot \hat{\mathbf{S}} + 2j\omega [\langle w_m \rangle - \langle w_e \rangle] = -\hat{P}_d$$

III. Transverse Electromagnetic Waves ($\rho_f = 0, \bar{\mathbf{J}} = 0$)

A. Wave equation

$$\nabla \times \bar{\mathbf{E}} = -\mu \frac{\partial \bar{\mathbf{H}}}{\partial t}$$

$$\nabla \times \bar{\mathbf{H}} = \epsilon \frac{\partial \bar{\mathbf{E}}}{\partial t}$$

$$\nabla \cdot \bar{\mathbf{E}} = 0$$

$$\nabla \cdot \bar{\mathbf{H}} = 0$$

$$\nabla \times (\nabla \times \bar{E}) = -\mu \frac{\partial}{\partial t} (\nabla \times \bar{H}) = -\mu \frac{\partial}{\partial t} \left(\epsilon \frac{\partial \bar{E}}{\partial t} \right)$$

$$\nabla \times (\nabla \times \bar{E}) = \nabla (\nabla \cdot \bar{E}) - \nabla^2 \bar{E} = -\epsilon \mu \frac{\partial^2 \bar{E}}{\partial t^2}$$

Wave equation

$$\nabla^2 \bar{E} = \frac{1}{c^2} \frac{\partial^2 \bar{E}}{\partial t^2}$$

where $c = \frac{1}{\sqrt{\epsilon\mu}}$ is the speed of the electromagnetic wave.

In free space $\mu = \mu_0 = 4\pi \times 10^{-7}$ henries/m and $\epsilon = \epsilon_0 \approx \frac{10^{-9}}{36\pi}$ farads/m, which leads to $c_0 = \frac{1}{\sqrt{\epsilon_0\mu_0}} \approx 3 \times 10^8$ m/s.

Similarly

$$\nabla \times (\nabla \times \bar{H}) = \nabla (\nabla \cdot \bar{H}) - \nabla^2 \bar{H} = \epsilon \frac{\partial}{\partial t} (\nabla \times \bar{E}) = -\epsilon \mu \frac{\partial^2 \bar{H}}{\partial t^2}$$

$$\nabla^2 \bar{H} = \frac{1}{c^2} \frac{\partial^2 \bar{H}}{\partial t^2}, c = \frac{1}{\sqrt{\epsilon\mu}}$$

B. Plane waves

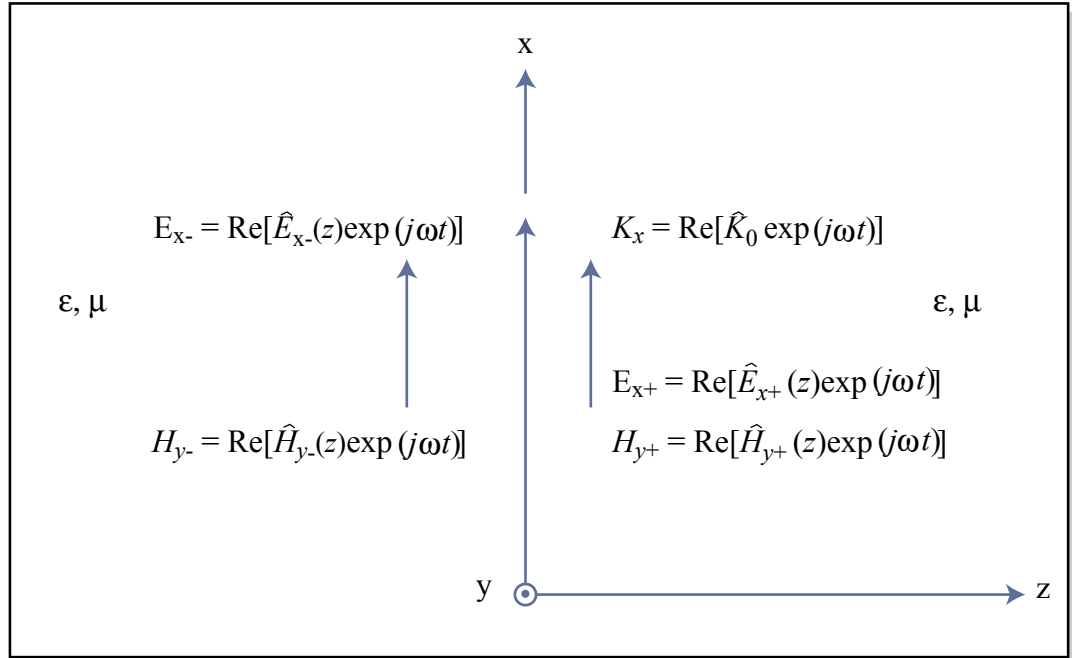


Image by MIT OpenCourseWare.

$$E_x(z, t) = \text{Re} \left[\hat{E}_x(z) e^{j\omega t} \right]$$

$$\frac{d^2 \hat{E}_x}{dz^2} = -\frac{\omega^2}{c^2} \hat{E}_x$$

$$\frac{d^2 \hat{E}_x}{dz^2} + k^2 \hat{E}_x = 0$$

where we have

$$k^2 = \frac{\omega^2}{c^2} = \omega^2 \epsilon \mu$$

$$k = \frac{2\pi}{\lambda} \text{ is the wavenumber, } \lambda \text{ is the wavelength}$$

$$k = \pm \frac{\omega}{c} \Rightarrow \omega = kc$$

$$\omega = 2\pi f = \frac{2\pi}{\lambda} c \Rightarrow f\lambda = c$$

$$\hat{E}_x = A_1 e^{jkz} + A_2 e^{-jkz}$$

$$E_x = \text{Re} \left[\underbrace{A_1 e^{j(\omega t + kz)}}_{\substack{\text{traveling wave} \\ \text{in the } -z \text{ direction}}} + \underbrace{A_2 e^{j(\omega t - kz)}}_{\substack{\text{traveling wave} \\ \text{in the } +z \text{ direction}}} \right]$$

For the wave in the $-z$ direction we have:

$$\omega t + kz = \text{constant}$$

$$\omega dt + kdz = 0$$

$$\frac{dz}{dt} = -\frac{\omega}{k} = -c$$

For the wave in the $+z$ direction we have:

$$\omega t - kz = \text{constant}$$

$$\omega dt - kdz = 0$$

$$\frac{dz}{dt} = \frac{\omega}{k} = +c$$

$$\hat{E}_x(z) = \begin{cases} \hat{E}_{x+} e^{-jkz} & z > 0 \\ \hat{E}_{x-} e^{+jkz} & z < 0 \end{cases}$$

$$\nabla \times \bar{E} = -\mu \frac{\partial \bar{H}}{\partial t} \Rightarrow \frac{d\hat{E}_x}{dz} = -j\omega\mu \hat{H}_y \Rightarrow \hat{H}_y = -\frac{1}{j\omega\mu} \frac{d\hat{E}_x}{dz}$$

$$\nabla \times \bar{H} = \epsilon \frac{\partial \bar{E}}{\partial t} \Rightarrow \frac{d\hat{H}_y}{dz} = -j\omega\epsilon \hat{E}_x$$

$$\hat{H}_y = \begin{cases} \frac{k}{\omega\mu} \hat{E}_{x+} e^{-jkz} & z > 0 \\ -\frac{k}{\omega\mu} \hat{E}_{x-} e^{+jkz} & z < 0 \end{cases}$$

$$\frac{k}{\omega\mu} = \frac{\omega\sqrt{\epsilon\mu}}{\omega\mu} = \sqrt{\frac{\epsilon}{\mu}}, \eta = \sqrt{\frac{\mu}{\epsilon}} \text{ is the wave impedance}$$

$$\hat{H}_y = \begin{cases} \frac{\hat{E}_{x+}}{\eta} e^{-jkz} & z > 0 \\ -\frac{\hat{E}_{x-}}{\eta} e^{+jkz} & z < 0 \end{cases}$$

Now we look at the boundary conditions:

$$E_x(z = 0_+, t) = E_x(z = 0_-, t) \Rightarrow \hat{E}_{x+} = \hat{E}_{x-}$$

$$H_y(z = 0_-, t) - H_y(z = 0_+, t) = K_x(z = 0, t) \Rightarrow \frac{-\hat{E}_{x-} - \hat{E}_{x+}}{\eta} = \hat{K}_0$$

$$\hat{E}_{x+} = \hat{E}_{x-} = -\frac{\eta \hat{K}_0}{2}$$

$$\hat{E}_x(z) = \begin{cases} -\frac{\eta \hat{K}_0}{2} e^{-jkz} & z > 0 \\ -\frac{\eta \hat{K}_0}{2} e^{jkz} & z < 0 \end{cases}$$

$$\hat{H}_y(z) = \begin{cases} -\frac{\hat{K}_0}{2} e^{-jkz} & z > 0 \\ \frac{\hat{K}_0}{2} e^{jkz} & z < 0 \end{cases}$$

$$\hat{S} = \frac{1}{2} \hat{E} \times \hat{H}^* = \begin{cases} \frac{\eta K_0^2}{8} \bar{i}_z & z > 0 \\ -\frac{\eta K_0^2}{8} \bar{i}_z & z < 0 \end{cases} \quad (\hat{K}_0 \text{ real})$$

$$E_x(z, t) = \text{Re} \left[\hat{E}_x(z) e^{j\omega t} \right] = \begin{cases} -\frac{\eta K_0}{2} \cos(\omega t - kz) & z > 0 \\ -\frac{\eta K_0}{2} \cos(\omega t + kz) & z < 0 \end{cases}$$

$$H_y(z, t) = \text{Re} \left[\hat{H}_y(z) e^{j\omega t} \right] = \begin{cases} -\frac{K_0}{2} \cos(\omega t - kz) & z > 0 \\ +\frac{K_0}{2} \cos(\omega t + kz) & z < 0 \end{cases}$$

$$S_z = E_x H_y = \begin{cases} \frac{\eta K_0^2}{4} \cos^2(\omega t - kz) & z > 0 \\ -\frac{\eta K_0^2}{4} \cos^2(\omega t + kz) & z < 0 \end{cases}$$

$$\langle S_z \rangle = \begin{cases} \frac{\eta K_0^2}{8} & z > 0 \\ -\frac{\eta K_0^2}{8} & z < 0 \end{cases}$$

C. Normal Incidence Onto a Perfect Conductor

$$\text{Incident Fields:} \quad \bar{E}_i(z, t) = \text{Re} \left[\hat{E}_i e^{j(\omega t - kz)} \bar{i}_x \right]$$

$$\bar{H}_i(z, t) = \text{Re} \left[\frac{\hat{E}_i}{\eta} e^{j(\omega t - kz)} \bar{i}_y \right]$$

$$\text{Reflected Fields:} \quad \bar{E}_r(z, t) = \text{Re} \left[\hat{E}_r e^{j(\omega t + kz)} \bar{i}_x \right]$$

$$\bar{H}_r(z, t) = \text{Re} \left[-\frac{\hat{E}_r}{\eta} e^{j(\omega t + kz)} \bar{i}_y \right]$$

$$k = \omega \sqrt{\epsilon \mu}, \eta = \sqrt{\frac{\mu}{\epsilon}}$$

The boundary conditions require that

$$E_x(z = 0, t) = E_{x,i}(z = 0, t) + E_{x,r}(z = 0, t) = 0$$

$$\hat{E}_i + \hat{E}_r = 0 \Rightarrow \hat{E}_r = -\hat{E}_i$$

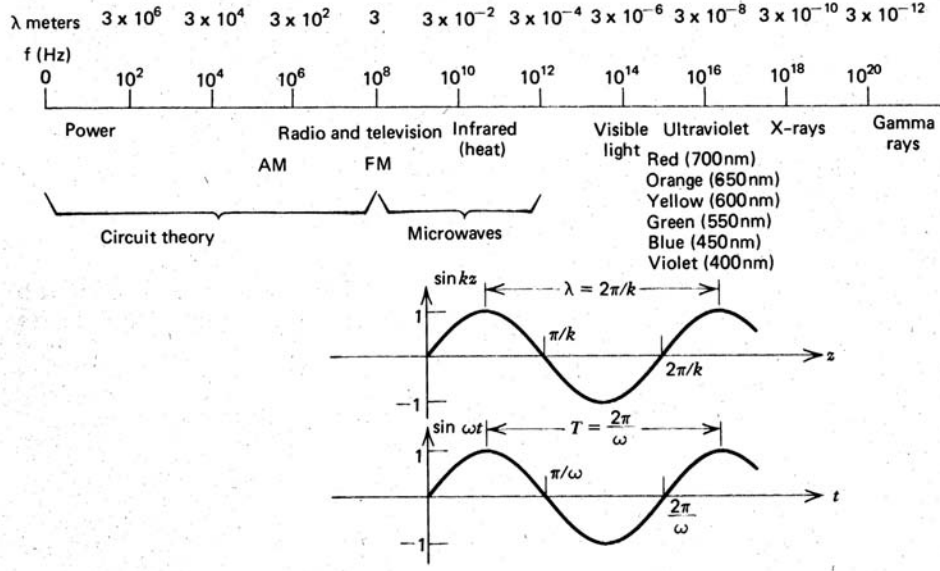


Figure 7-7 Time varying electromagnetic phenomena differ only in the scaling of time (frequency) and size (wavelength). In linear dielectric media the frequency and wavelength are related as $f\lambda = c$ ($\omega = kc$), where $c = 1/\sqrt{\epsilon\mu}$ is the speed of light in the medium.

From *Electromagnetic Field Theory: A Problem Solving Approach*, by Markus Zahn, 1987. Used with permission.

For $\hat{E}_i = E_i$ real we have:

$$E_x(z, t) = E_{x,i}(z, t) + E_{x,r}(z, t) = \text{Re} \left[\hat{E}_i \left(e^{-jkz} - e^{+jkz} \right) e^{j\omega t} \right]$$

$$= 2E_i \sin(kz) \sin(\omega t)$$

$$H_y(z, t) = H_{y,i}(z, t) + H_{y,r}(z, t) = \text{Re} \left[\frac{\hat{E}_i}{\eta} \left(e^{-jkz} + e^{+jkz} \right) e^{j\omega t} \right]$$

$$= \frac{2E_i}{\eta} \cos(kz) \cos(\omega t)$$

$$K_z(z = 0, t) = H_y(z = 0, t) = \frac{2E_i}{\eta} \cos(\omega t)$$

Radiation pressure in free space ($\mu = \mu_0, \epsilon = \epsilon_0$)

$$\frac{\text{Force}_z}{\text{Area}} \Big|_{z=0} = \frac{1}{2} \bar{K} \times \mu_0 \bar{H} = \frac{1}{2} \mu_0 K_x H_y \Big|_{z=0} \bar{i}_z = \frac{1}{2} \mu_0 H_y^2(z = 0) \bar{i}_z$$

$$= \frac{2\mu_0 E_i^2}{\eta_0^2} \cos^2(\omega t) \bar{i}_z$$

$$= \frac{2\mu_0}{\mu_0/\epsilon_0} E_i^2 \cos^2(\omega t) \bar{i}_z$$

$$= 2\epsilon_0 E_i^2 \cos^2(\omega t) \bar{i}_z$$

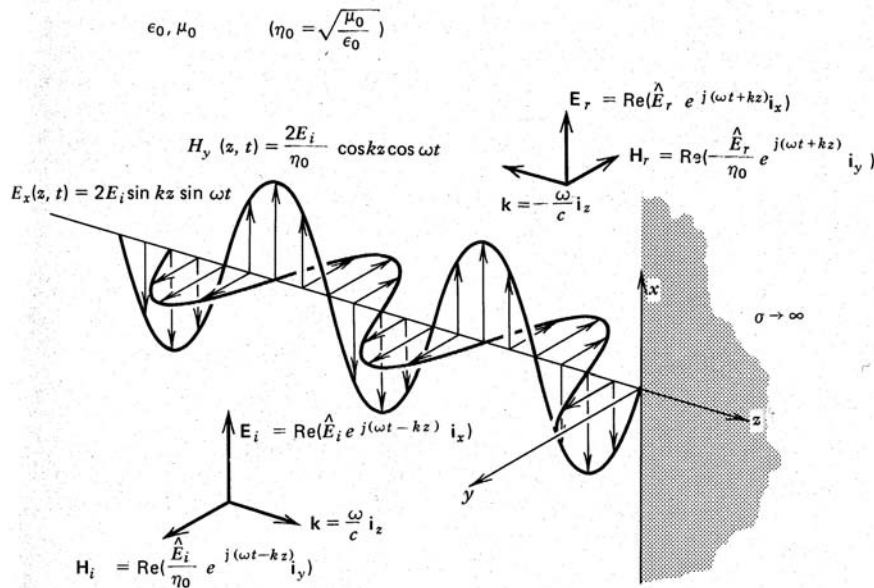


Figure 7-13 A uniform plane wave normally incident upon a perfect conductor has zero electric field at the conducting surface thus requiring a reflected wave. The source of this reflected wave is the surface current at $z=0$, which equals the magnetic field there. The total electric and magnetic fields are 90° out of phase in time and space.

From *Electromagnetic Field Theory: A Problem Solving Approach*, by Markus Zahn, 1987. Used with permission.

IV. Normal Incidence Onto a Dielectric

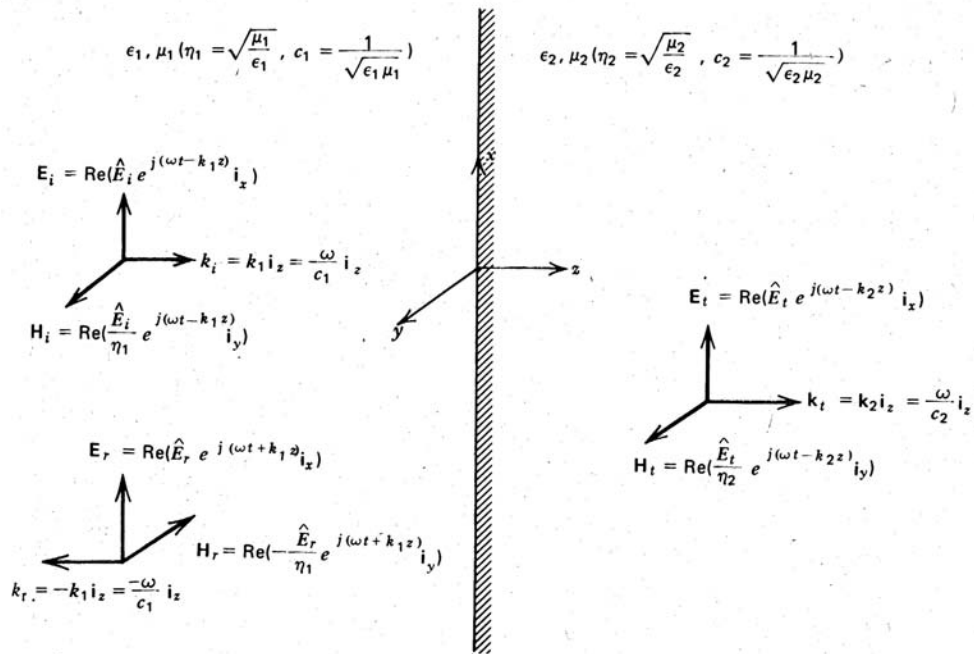


Figure 7-14 A uniform plane wave normally incident upon a dielectric interface separating two different materials has part of its power reflected and part transmitted.

From *Electromagnetic Field Theory: A Problem Solving Approach*, by Markus Zahn, 1987. Used with permission.

$$\begin{aligned}
\bar{E}_i(z, t) &= \text{Re} \left[\hat{E}_i e^{j(\omega t - k_1 z)} \bar{i}_x \right], k_1 = \omega \sqrt{\epsilon_1 \mu_1} \\
\bar{H}_i(z, t) &= \text{Re} \left[\frac{\hat{E}_i}{\eta_1} e^{j(\omega t - k_1 z)} \bar{i}_y \right], \eta_1 = \sqrt{\mu_1 / \epsilon_1} \\
\bar{E}_r(z, t) &= \text{Re} \left[\hat{E}_r e^{j(\omega t + k_1 z)} \bar{i}_x \right] \\
\bar{H}_r(z, t) &= \text{Re} \left[\frac{-\hat{E}_r}{\eta_1} e^{j(\omega t + k_1 z)} \bar{i}_y \right] \\
\bar{E}_t(z, t) &= \text{Re} \left[\hat{E}_t e^{j(\omega t - k_2 z)} \bar{i}_x \right], k_2 = \omega \sqrt{\epsilon_2 \mu_2} \\
\bar{H}_t(z, t) &= \text{Re} \left[\frac{\hat{E}_t}{\eta_2} e^{j(\omega t - k_2 z)} \bar{i}_y \right], \eta_2 = \sqrt{\mu_2 / \epsilon_2}
\end{aligned}$$

$$\begin{aligned}
E_x(z = 0_-) = E_x(z = 0_+) &\Rightarrow \hat{E}_i + \hat{E}_r = \hat{E}_t \\
H_y(z = 0_-) = H_y(z = 0_+) &\Rightarrow \frac{\hat{E}_i - \hat{E}_r}{\eta_1} = \frac{\hat{E}_t}{\eta_2} \\
R &\equiv \frac{\hat{E}_r}{\hat{E}_i} = \frac{\eta_2 - \eta_1}{\eta_1 + \eta_2} \text{ is the Reflection coefficient} \\
T &\equiv \frac{\hat{E}_t}{\hat{E}_i} = \frac{2\eta_2}{\eta_1 + \eta_2} \text{ is the Transmission coefficient} \\
1 + R &= T
\end{aligned}$$

$$\begin{aligned}
\langle S_{z,i} \rangle &= \frac{1}{2} \text{Re} \left[\hat{E}_x(z) \hat{H}_y^*(z) \right] \\
&= \frac{1}{2\eta_1} \text{Re} \left[\left(\hat{E}_i e^{-jk_1 z} + \hat{E}_r e^{jk_1 z} \right) \left(\hat{E}_i^* e^{+jk_1 z} - \hat{E}_r^* e^{-jk_1 z} \right) \right] \\
&= \frac{1}{2\eta_1} \left[|\hat{E}_i|^2 - |\hat{E}_r|^2 \right] \\
&\quad + \frac{1}{2\eta_1} \underbrace{\text{Re} \left[\hat{E}_r \hat{E}_i^* e^{2jk_1 z} - \hat{E}_r^* \hat{E}_i e^{-2jk_1 z} \right]}_{\text{pure imaginary}} \\
&= \frac{1}{2\eta_1} \left[|\hat{E}_i|^2 - |\hat{E}_r|^2 \right] \\
&= \frac{|\hat{E}_i|^2}{2\eta_1} [1 - R^2] \\
\langle S_{z,t} \rangle &= \frac{1}{2\eta_2} |\hat{E}_t|^2 = \frac{|\hat{E}_i|^2 T^2}{2\eta_2} = \frac{|\hat{E}_i|^2 (1 - R^2)}{2\eta_1} = \langle S_{z,i} \rangle
\end{aligned}$$

V. Lossy Dielectrics - $\bar{J} = \sigma \bar{E}$

$$\text{Ampere's Law: } \nabla \times \bar{H} = \bar{J} + \epsilon \frac{\partial \bar{E}}{\partial t} = \sigma \bar{E} + \epsilon \frac{\partial \bar{E}}{\partial t}$$

For $e^{j\omega t}$ fields

$$\nabla \times \hat{H} = (j\omega\epsilon + \sigma) \hat{E} = j\omega\epsilon \left(1 + \frac{\sigma}{j\omega\epsilon} \right) \hat{E}$$

Define complex permittivity by $\hat{\epsilon} = \epsilon \left(1 + \frac{\sigma}{j\omega\epsilon}\right)$. Then complex amplitude solutions are the same as real amplitude solutions if we replace ϵ by $\hat{\epsilon}$:

$$k = \omega\sqrt{\hat{\epsilon}\mu}, \quad \eta = \sqrt{\frac{\mu}{\hat{\epsilon}}} = \sqrt{\frac{\mu}{\epsilon\left(1 + \frac{\sigma}{j\omega\epsilon}\right)}}$$

$$k = \omega\sqrt{\epsilon\mu\left(1 + \frac{\sigma}{j\omega\epsilon}\right)}$$

A. Low loss limit: $\frac{\sigma}{\omega\epsilon} \ll 1$

$$\begin{aligned} k &= \omega\sqrt{\epsilon\mu}\sqrt{1 + \frac{\sigma}{j\omega\epsilon}} \\ &\approx \omega\sqrt{\epsilon\mu}\left(1 + \frac{1}{2}\frac{\sigma}{j\omega\epsilon}\right) \\ &\approx \omega\sqrt{\epsilon\mu} - \frac{j\sigma}{2}\sqrt{\frac{\mu}{\epsilon}} \\ &\approx \omega\sqrt{\epsilon\mu} - j\frac{\sigma\eta}{2} \\ e^{-jkz} &= e^{-j\omega\sqrt{\epsilon\mu}z} e^{-j(-j\frac{\sigma\eta}{2})z} \\ &= e^{-j\omega\sqrt{\epsilon\mu}z} \underbrace{e^{-\frac{\sigma\eta}{2}z}}_{\text{slow exponential decay}} \end{aligned}$$

B. Large loss limit: $\frac{\sigma}{\omega\epsilon} \gg 1$

$$\begin{aligned} k &= \omega\sqrt{\epsilon\mu}\sqrt{1 + \frac{\sigma}{j\omega\epsilon}} \\ &\approx \omega\sqrt{\epsilon\mu}\sqrt{\frac{\sigma}{\omega\epsilon}} \underbrace{\sqrt{-j}}_{\frac{1-j}{\sqrt{2}}} \\ &\approx \sqrt{\frac{\omega\mu\sigma}{2}}(1-j) \\ &\approx \frac{1-j}{\delta} \quad \delta \equiv \sqrt{\frac{2}{\omega\mu\sigma}} \text{ is the skin depth} \\ e^{-jkz} &= e^{-j(1-j)z/\delta} = e^{-jz/\delta} \underbrace{e^{-z/\delta}}_{\text{fast exponential decay}} \end{aligned}$$